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MICROELECTRONIC ENGINEERING**

SPICE Model Parameters for RIT MOSFET's

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OUTLINE

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SPICE Level 1 Model

SPICE Level 3 Model

BSIM3 Model

SPICE Parameter Calculator

SPICE Parameters for RIT MOSFETs

WinSpice

Examples

Parameter Extraction Using UTMOST

ATHENA > ATLAS > UTMOST > SPICE

References

SPICE - SIMULATION PROGRAM FOR INTEGRATED CIRCUIT ENGINEERING

Device models used by SPICE (Simulation Program for Integrated Circuit Engineering) simulators can be divided into three classes: First Generation Models (Level 1, Level 2, Level 3 Models), Second Generation Models (BISM, HSPICE Level 28, BSIM2) and Third Generation Models (BSIM3, Level 7, Level 48, etc.) The newer generations can do a better job with short channel effects, local stress, transistors operating in the sub-threshold region, gate leakage (tunneling), noise calculations, temperature variations and the equations used are better with respect to convergence during circuit simulation.

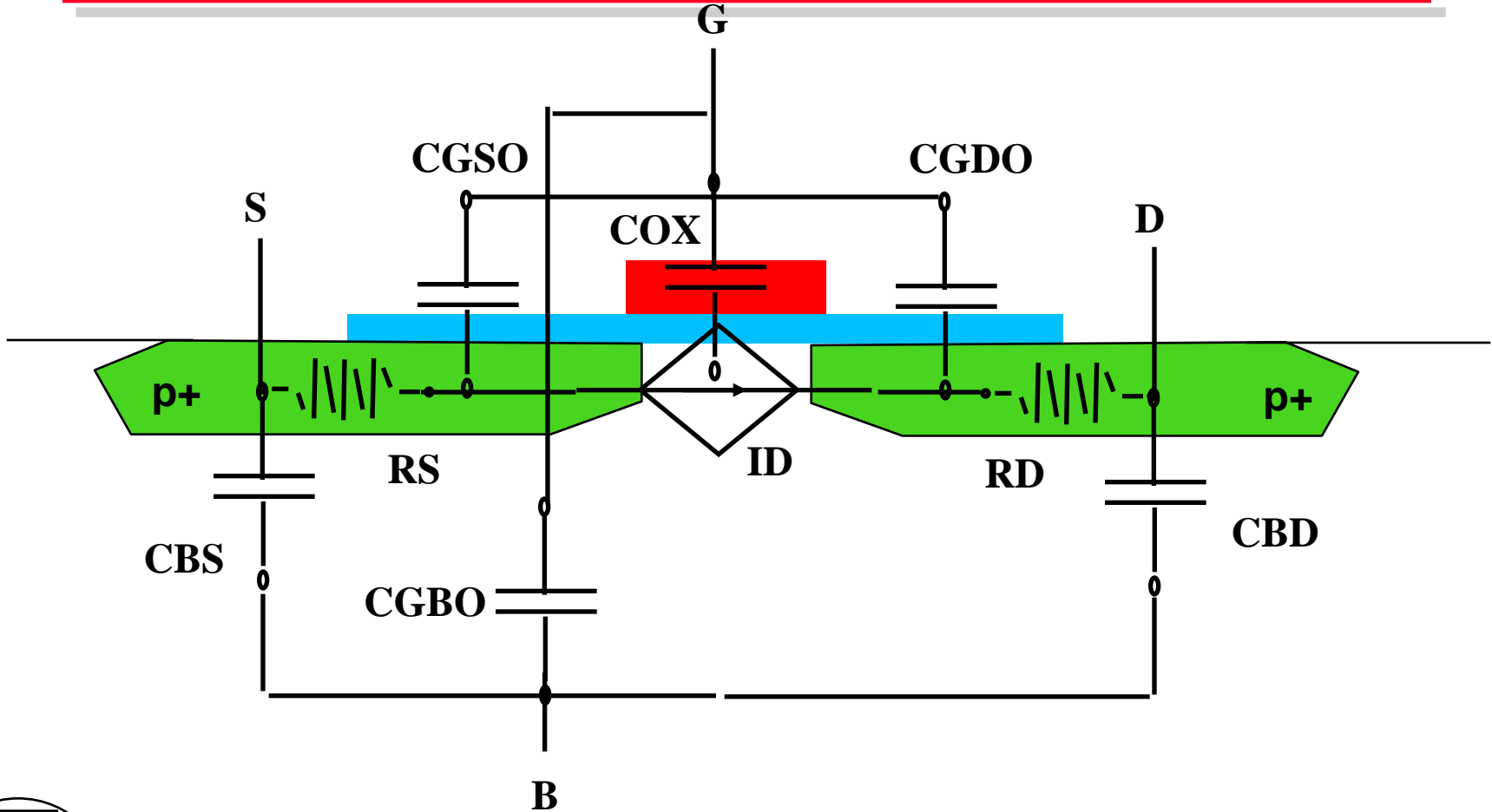
SPICE LEVEL-1 SHICHMAN AND HODGES

If we understand the Level 1 model we can better understand the other models. The Level 1 model by Shichman and Hodges uses basic device physics equations for MOSFET threshold voltage and drain current in the saturation and non-saturation regions of operation. Mobility is assumed to be a function of total doping concentration only and a parameter called LAMBDA is used to model channel length modulation.



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SPICE LEVEL-1 MOSFET MODEL



where ID is a dependent current source using the equations on the next page

SPICE LEVEL-1 EQUATIONS FOR U_0 , V_{to} AND I_D

Mobility:

$$\mu = \mu_{\min} + \frac{(\mu_{\max} - \mu_{\min})}{\{1 + (N/N_{\text{ref}})^\alpha\}}$$

Parameter	Arsenic	Phosphorous	Boron
μ_{\min}	52.2	68.5	44.9
μ_{\max}	1417	1414	470.5
N_{ref}	9.68×10^{16}	9.20×10^{16}	2.23×10^{17}
α	0.680	0.711	0.719

Threshold Voltage:

+/-

nmos/pmos

$$V_{TO} = \Phi_{ms} - q \text{NSS}/C_{ox}' \pm [-2[\Phi F] \pm 2(q\epsilon_s \text{NSUB} [\Phi F])^{0.5}/C_{ox}']$$

$$[\Phi F] = (KT/q) \ln(\text{NSUB}/n_i) \quad \text{where } n_i = 1.45E10 \text{ and } KT/q = 0.026$$

Absolute value

Drain Current:

Non-Saturation

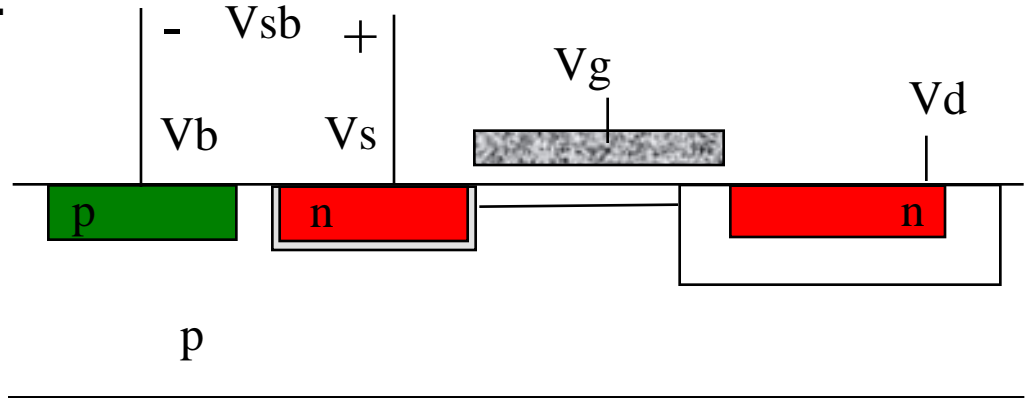
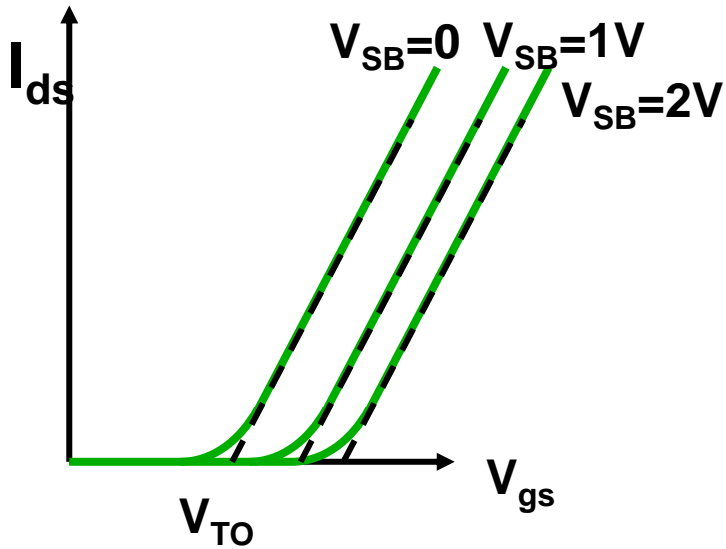
$$I_D = \frac{\mu W C_{ox}' (V_g - V_t - V_d/2) V_d (1 + \lambda V_{ds})}{L}$$

Saturation

$$I_{D\text{sat}} = \frac{\mu W C_{ox}' (V_g - V_t)^2 (1 + \lambda V_{ds})}{2L}$$

BACK-BIASING EFFECTS – GAMMA

Body Effect coefficient GAMMA or γ :



$$\gamma = \frac{1}{C_{ox}'} \sqrt{2q\epsilon_{si} N_{sub}}$$

$$V_T = \Phi_{MS} - \frac{Q_{ss}}{C_{ox}'} + 2\phi_F + \gamma \sqrt{2\phi_F + V_{SB}}$$

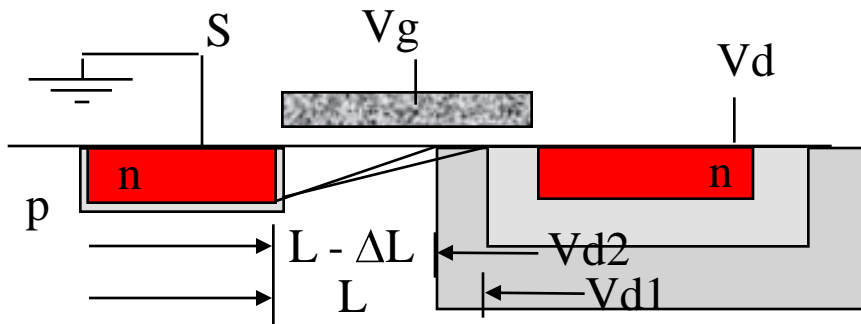
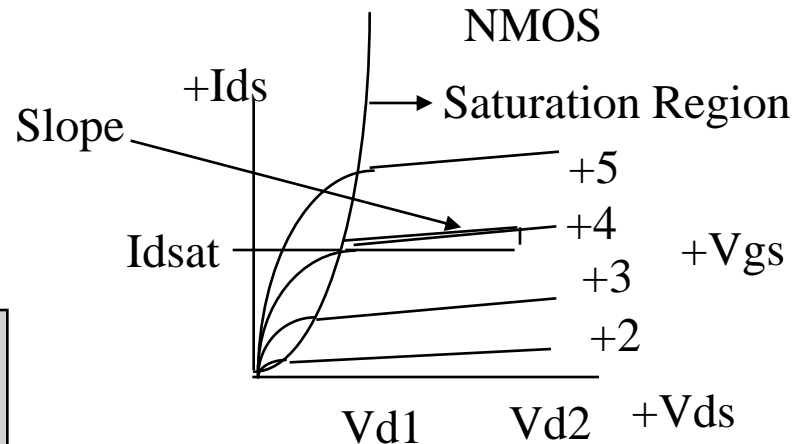
where $\epsilon_{si} = 11.7$ and $\epsilon_{ox} = 3.9$
 $\epsilon_0 = 8.8eE-14F/cm$
 $q = 1.6E-19$

CHANNEL LENGTH MODULATION - LAMBDA

Channel Length Modulation

Parameter λ

$\lambda = \text{Slope} / I_{\text{dsat}}$



$$I_{D_{\text{sat}}} = \frac{\mu W C_{\text{ox}}'}{2L} (V_g - V_t)^2 (1 + \lambda V_{\text{ds}})$$

Saturation Region

$$I_D = \frac{\mu W C_{\text{ox}}'}{L} (V_g - V_t - V_d/2) V_d (1 + \lambda V_{\text{ds}})$$

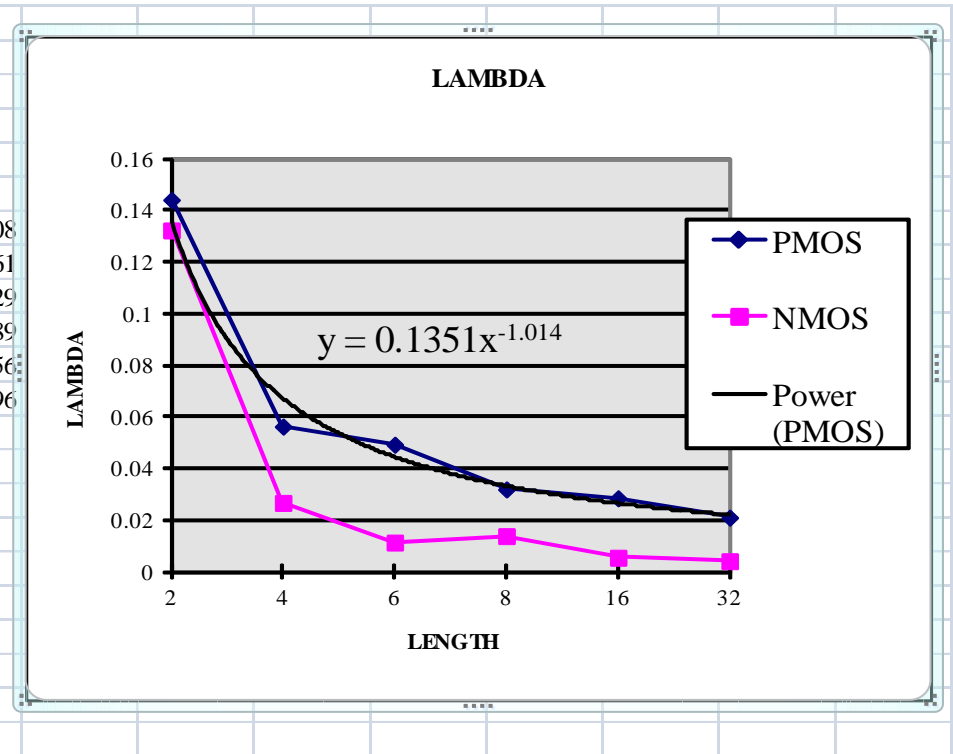
Non Saturation Region

NMOS Transistor

DC Model, λ is the channel length modulation parameter and is different for each channel length, L. Typical value might be 0.02

LAMBDA VERSUS CHANNEL LENGTH

μA	SLOPE	IDSAT	W	L	LAMBDA	
UNIT					PMOS	NMOS
205	4.9	6.8	32	2	0.144118	0.132308
71	2	7.1	32	4	0.056338	0.026761
56	1.8	7.3	32	6	0.049315	0.011429
34	1.2	7.5	32	8	0.032	0.013889
21	1	7	32	16	0.028571	0.005556
8.8	0.8	7.6	32	32	0.021053	0.004196
415	4.3	6.5	32	2	0.132308	
137	0.95	7.1	32	4	0.026761	
91	0.4	7	32	6	0.011429	
137	0.5	7.2	32	8	0.013889	
27	0.2	7.2	32	16	0.005556	
15	0.15	7.15	32	32	0.004196	



Need different model for each different length transistor

SPICE LEVEL-1 PARAMETERS

SPICE LEVEL 1 MODEL FOR MOS TRANSISTORS:

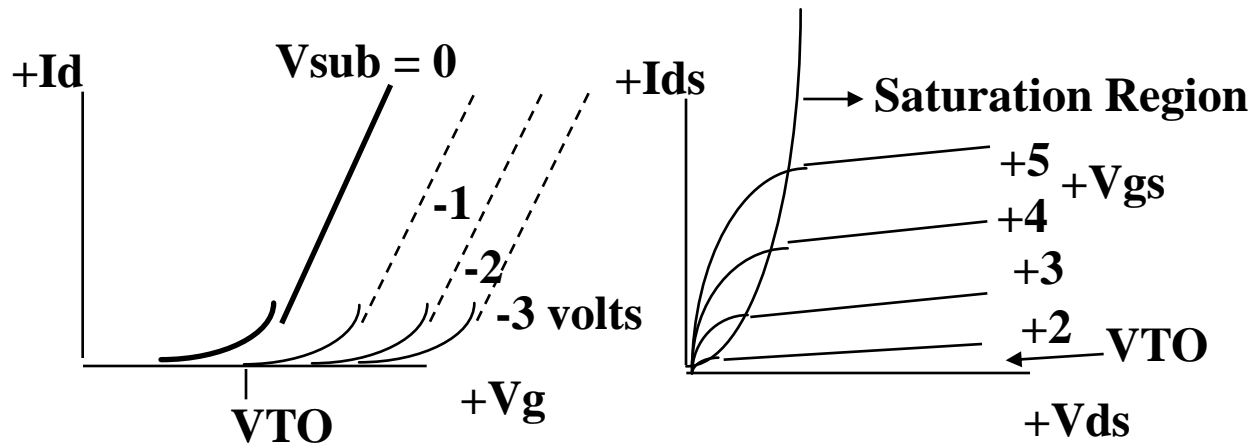
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|------------|---------|----------|----------|---------|
| 1. LEVEL=1 | 7. RD | 13. CGSO | 19. CJSW | 25. NFS |
| 2. VTO | 8. RS | 14. CGDO | 20. MJSW | 26. TPG |
| 3. KP | 9. CBD | 15. CGBO | 21. JS | 27. XJ |
| 4. GAMMA | 10. CBS | 16. RSH | 22. TOX | 28. LD |
| 5. PHI | 11. IS | 17. CJ | 23. NSUB | 29. UO |
| 6. LAMBDA | 12. PB | 18. MJ | 24. NSS | |
- 30.-41. PARAMETERS FOR SHORT CHANNEL AND NOISE (Use Defaults)

SPICE is not case sensitive.

The O in these parameters is “oh” not zero.

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

1. LEVEL=1 Shichman-Hodges Model (If not specified the default is LEVEL=1)
2. VTO zero bias threshold voltage (Do not use, let SPICE calculate from Nsub,TOX unless an VT adjust ion implant is used to set VTO at some value)



3. KP transconductance parameter (Do not use, let SPICE calculate from UO, COX')

$$KP = UO COX' = UO \epsilon_r \epsilon_o / TOX$$

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SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

4. GAMMA bulk threshold parameter (Do not use, let SPICE calculate from NSUB,COX')

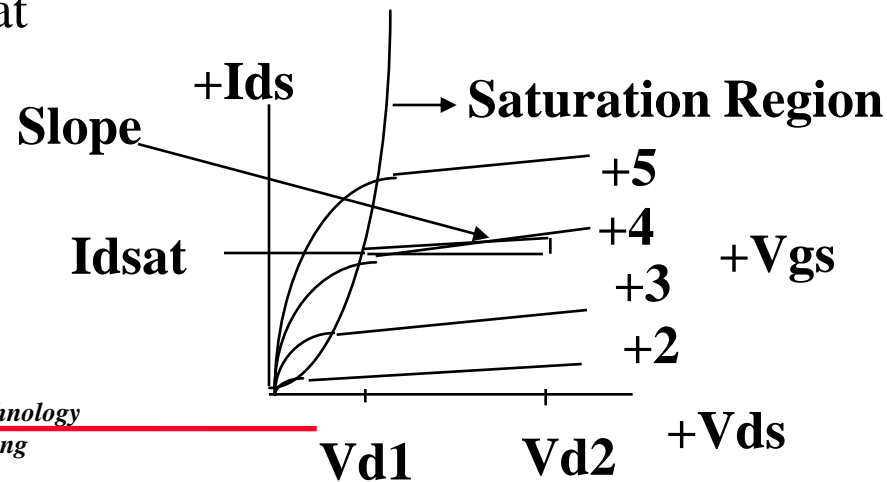
$$\text{GAMMA} = [2q \epsilon_{\text{si}} \epsilon_0 \text{NSUB}/C'_{\text{ox}}]^2 \quad \text{where } \epsilon_{\text{si}} \epsilon_0 = (11.7)(8.85\text{E-}12) \\ \text{and } q = 1.6\text{E-}19$$

5. PHI is the surface inversion potential, 2 x Intrinsic Level to Fermi Level difference in Volts (Do not use, let SPICE calculate from NSUB)

$$\text{PHI} = 2[\Phi_F] = 2 (KT/q) \ln (\text{NSUB}/n_i) \quad \text{where } KT/q = 0.026 \\ n_i = 1.45\text{E}10$$

6. LAMBDA is the channel length modulation parameter, Slope in saturation region divided by Idsat

$$\lambda = \text{Slope} / \text{Idsat}$$



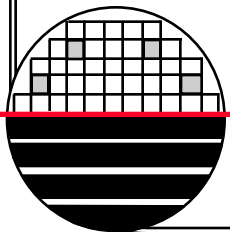
SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

7. RD the series drain resistance can either be given as a resistance value or through RSH the drain/source sheet resistance and the number of squares NRS. NRS: Is from the device layout. RSH: Is measured by four point probe or Van Der Pauw structures

(Do not use, let SPICE calculate from sheet resistance, RSH, and number of squares in drain, NRD)

8. RS is the series source resistance can either be given as a resistance value or through RSH the drain/source sheet resistance and the number of squares NRS. NRS: Is from the device layout. RSH: Is measured by four point probe or Van Der Pauw structures

(Do not use, let SPICE calculate from sheet resistance, RSH, and number of squares in source, NRS)



SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

9. CBD zero bias bulk to drain junction capacitance (Do not use, let SPICE calculate from CJ and CJSW and AD (Area of Drain) and PD (Perimeter of Drain))

$$CBD = CJ AD + CJSW PD$$

10. CBS zero bias bulk to source junction capacitance (Do not use, let SPICE calculate from CJ and CJSW and AS (Area of Source) and PD (Perimeter of Source))

$$CBS = CJ AS + CJSW PS$$

11. IS is the bulk junction saturation current in the ideal diode equation.

$$I = IS (\exp^{qVA/KT} - 1)$$

- (Do not use, let SPICE calculate from JS and AD (Area of Drain) and AS (Area of Source))

$$IS = JS (AD + AS)$$

12. PB is the junction built in voltage

$$PB = (KT/q)\ln (NSUB/ni) + 0.56 \text{ or } PB = (KT/q) \ln (Na Nd/ni^2)$$

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

$$C_{ox}' = \epsilon_r \epsilon_0 / TOX = 3.9 \epsilon_0 / TOX$$

13. CGSO is the gate-to-source overlap capacitance (per meter channel width)

$$CGSO = C_{ox}' (\text{mask overlap in L direction} + LD) \quad \text{F/m}$$

14. CGDO is the gate-to-drain overlap capacitance (per meter channel width)

$$CGDO = C_{ox}' (\text{mask overlap in L direction} + LD) \quad \text{F/m}$$

15. CGBO is the gate-to-bulk overlap capacitance (per meter channel length)

$$CGBO = C_{\text{field_oxide}} * \text{mask overlap in W direction} \quad \text{F/m}$$

$$C_{\text{field_oxide}} = \epsilon_r \epsilon_0 / X_{\text{FieldOX}}$$

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

16. RSH is the drain and source diffusion sheet resistance. Measured from four point probe or Van Der Pauw structures.
17. CJ is the zero bias bulk junction bottom capacitance per square meter of junction area. $CJ = \epsilon_r \epsilon_0 / W$ where W is width of space charge layer.

$$CJ = \epsilon_r \epsilon_0 [2\epsilon_r \epsilon_0 (\Psi_0 - VA) / q N_{sub}]^{-m} \quad \text{F/m}^2$$

$$\text{where } \Psi_0 = PB = (KT/q) \ln (N_{SUB}/n_i) + 0.56$$

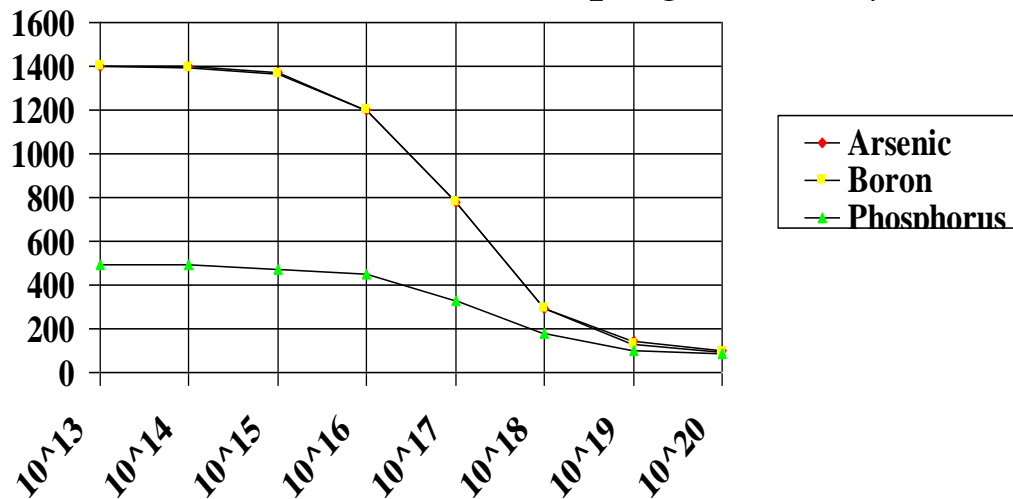
m = junction grading coefficient = 0.5

18. MJ is the junction grading coefficient = 0.5
19. CJSW is the zero bias bulk junction sidewall capacitance per meter of junction perimeter. $CJSW = CJ XJ$
20. MJSW is the junction grading coefficient = 0.5
21. JS is the bulk junction saturation current density in Amperes per square meter

$$JS = q n_i^2 (D_p / N_d L_p + D_n / N_a L_n) \text{ where } D = (KT/q) \mu \text{ and } L = (D\tau)^{0.5}$$

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

- 22. TOX is the gate oxide thickness, measured by ellipsometer or reflectance spectroscopy (Nanospec).
- 23. NSUB the substrate doping is given by the wafer manufacturer or measured by four point probe technique. In both cases NSUB is given indirectly by the resistivity, Rho. $Rho = 1/(q\mu(N)N)$ where $q = 1.6E-19$ coul, N is the substrate doping NSUB, $\mu(N)$ is the mobility, a function of N.



Emperical Equation:

$$\mu = \mu_{min} + \frac{\mu_{max} - \mu_{min}}{\{1 + (N/N_{ref})^\alpha\}}$$

	Electrons	Holes
μ_{min}	92	47.7
μ_{max}	1360	495
N_{ref}	1.3E17	6.3E16
α	0.91	0.76

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

24. NSS: The surface state density is a parameter used in the calculation of the zero-bias threshold voltage (ie. $V_{source} = V_{substrate}$), V_{T0} is obtained from transistor curves.

$$V_{T0} = \Phi_{ms} - q \text{NSS}/C_{ox}' - 2 \Phi_F - 2 (q \epsilon_s N_{SUB} \Phi_F)^{0.5}/C_{ox}'$$

$$\Phi_F = (KT/q) \ln (N_{SUB}/n_i) \quad \text{where } n_i = 1.45E10 \text{ and } KT/q = 0.026$$

$$\Phi_{ms} = \Phi_m - (X + E_g/2 - \Phi_F) \quad \text{where } \Phi_m = \text{gate work function}$$

$$X = 4.15 \text{ eV}, E_g = 1.12 \text{ eV}$$

$$\epsilon_s = \epsilon_r \epsilon_o = 11.7 \epsilon_o$$

Since everything is known

$$C_{ox}' = \epsilon_r \epsilon_o / TOX = 3.9 \epsilon_o / TOX$$

in equations above, NSS
can be calculated

25. NFS is the fast surface state density, usually left at zero.

SPICE LEVEL-1 PARAMETERS FOR MOSFET (cont.)

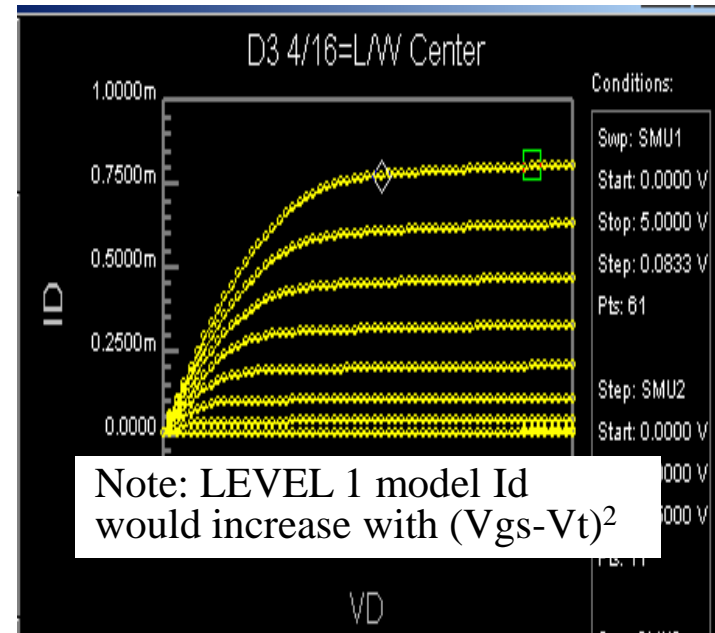
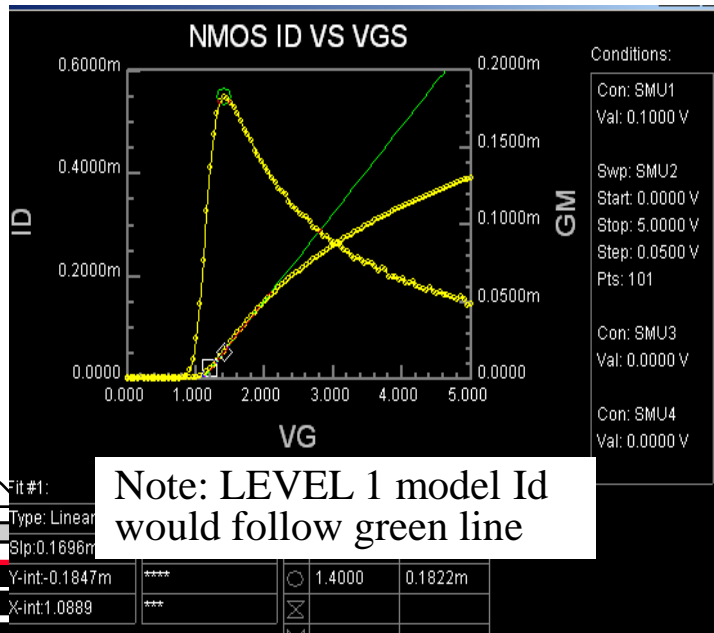
26. TPG is the type of gate. for aluminum TPG=0, for n+ poly TPG = 1,
for p+ poly TPG= -1
27. XJ metallurgical junction depth, measured by groove and stain techniques.
28. LD lateral diffusion distance, inferred from process knowledge
29. UO is the surface mobility taken as 1/2 the bulk mobility or
extracted to give correct Id value on measured Id vs Vds characteristics
in the saturation region. For best results make measurements on a transistor with
large channel length so that λ is small and the lateral diffusion can be neglected.

$$I_{Dsat} = \frac{\mu W C_{ox}' (V_g - V_t)^2 (1 + \lambda V_{ds})}{2L}$$

30. - 41. Parameters associated with short channel devices and noise in
MOSFETs

SPICE 2ND GENERATION MODELS AND PARAMETERS

2ND generation MOSFET models improve over the Level 1 models because they model sub-threshold current, mobility as a function of vertical and lateral electric field strength, threshold voltage reduction as a function drain voltage or drain induced barrier lowering (DIBL). This model has separate equations for drain current for different regions of operation. The discontinuity at the transition points can make problems in program convergence during circuit simulation.



TERADA-MUTA METHOD FOR EXTRACTING L_{eff} and R_{ds}

Terada-Muta Method for L_{eff} and R_{ds}

In the linear region (V_D is small):

$$I_D = \frac{\mu W C_{ox}' (V_g - V_t - V_D/2)}{L_{eff}} V_D$$

$$L_{eff} = L_m - \Delta L$$

$$I_D = 1/R_m V_D$$

where ΔL is correction due to processing

L_m is the mask length

$$R_m = V_D/I_D = \text{measured resistance}$$

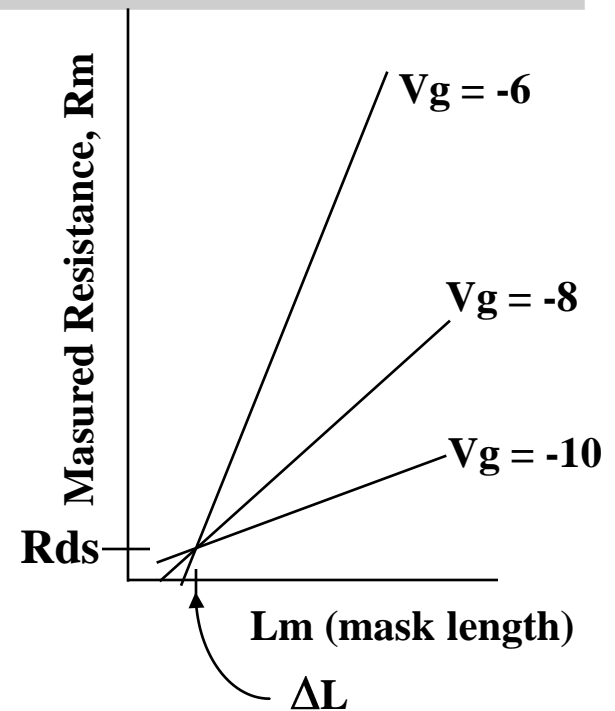
$$= L_m / (\mu W C_{ox}' (V_g - V_t)) - \Delta L / \mu W C_{ox}' (V_g - V_t)$$

so measure R_m for different channel length transistors and plot R_m vs L_m

where R_m = intersect find value for ΔL and R_{ds}

Then L_{eff} can be calculated for each different length transistor

$$\text{from } L_{eff} = L_m - \Delta L$$



2nd GENERATION MODELS EQUATIONS FOR MOBILITY

The mobility used in the equations for I_{ds} is the effective mobility, U_{eff} . Starting with U_0 from level 1, U_{eff} is found. The parameter THETA is introduced to model mobility degradation due to high vertical electric fields (larger values of $V_{gs} - V_{TO}$).

$$U_{eff}^* = \frac{U_0}{(1 + THETA (V_{gs} - V_{TO}))}$$

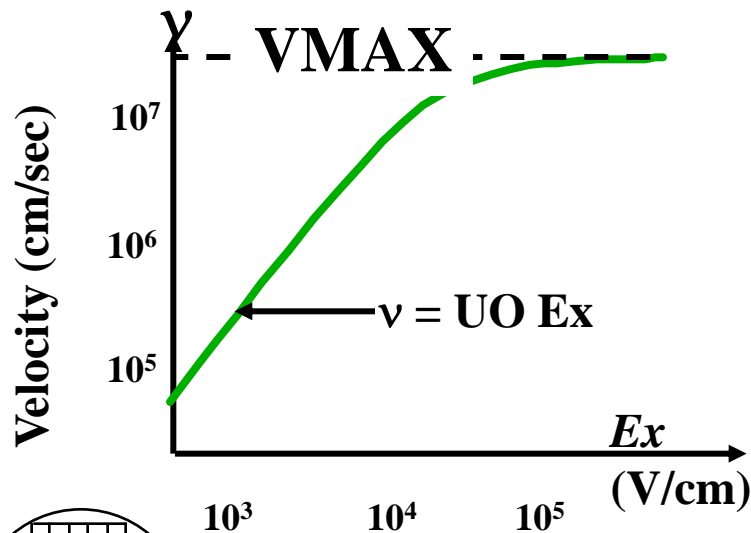
Measure I_{ds} for a wide transistor with low value of V_{ds} and large value of V_{gs} and using L_{eff} from Terata-Muta method and LAMBDA from level 1, calculate THETA from these two equations.

$$I_{dsat} = \frac{U_{eff} W Cox'}{2L_{eff}} (V_g - V_t)^2 (1 + \lambda V_{ds})$$

Warning: Curvature also due to R_{DS} so V_{ds} is $(V_{applied} - R_{ds} * I_{dsat})$ requires an iterative approach to find THETA

2ND GENERATION EQUATIONS FOR MOBILITY (cont.)

The parameter VMAX is introduced to model the decrease in mobility at higher Vds due to velocity saturation. Ideally, carrier velocity is directly proportional to the applied electric field. However, at very high lateral electric fields, Ex, this relationship ceases to be accurate - the carrier velocity saturates at VMAX.



$$U_{\text{eff}} = \frac{UO}{\left((1 + \text{THETA} (V_{\text{gs}} - V_{\text{TO}})) \left(1 + UO \frac{v_{\text{de}}}{V_{\text{MAX}} L_{\text{eff}}} \right) \right)}$$

Where, $V_{\text{de}} = \min (V_{\text{ds}}, V_{\text{dsat}})$

Note: other models (equations) for mobility exist and use parameters such as UCRIT, UEXP, ULTRA, ECRIT, ESAT

2ND GENERATION MODEL EQUATIONS FOR THRESHOLD VOLTAGE

The parameter ETA is used to describe DIBL (Drain Induced Barrier Lowering) resulting in a modification to the LEVEL 1 equation for threshold voltage.

$$V_{T0} = \Phi_{ms} - \phi_{ETA} - q N_{SS}/C_{ox}' - 2 \Phi_F - 2 (q\epsilon_s N_{SUB} \Phi_F)^{0.5}/C_{ox}'$$

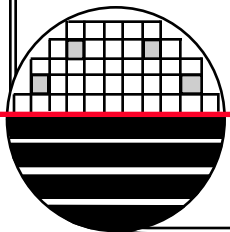
$$\phi_{ETA} = \frac{(-8.14E-22)*ETA}{C_{ox}'L_{eff}^3} V_{ds}$$

2ND GENERATION EQUATIONS FOR NARROW WIDTH

DELTA is introduced to model narrow channel effects on threshold voltage. The parameter WD (channel width reduction from drawn value) is used to calculate the effective channel width. DELTA is used in the calculation of threshold voltage.

$$\text{DELTA} = \frac{q \text{ NSUB } X_{ds}^2}{\epsilon_0 \epsilon_{si} 2 \text{ PHI}}$$

Note: a dimensionless number typically ~3

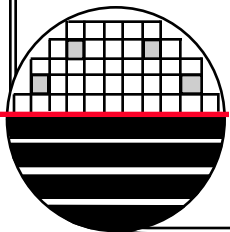
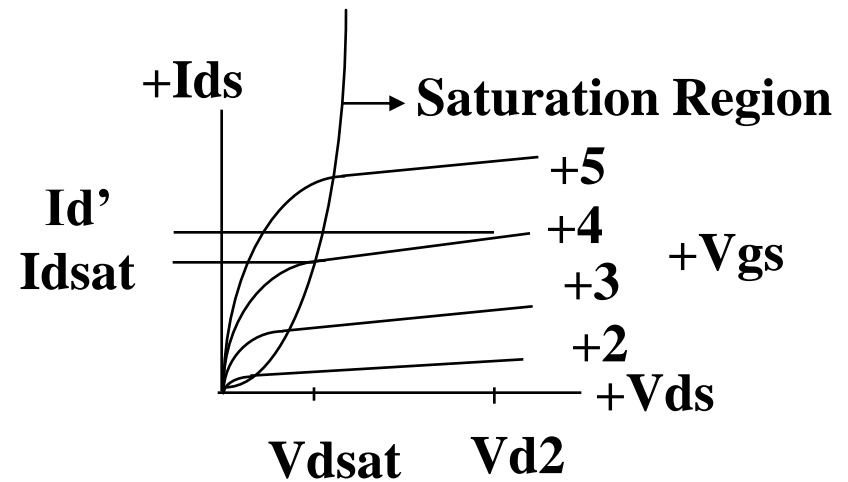
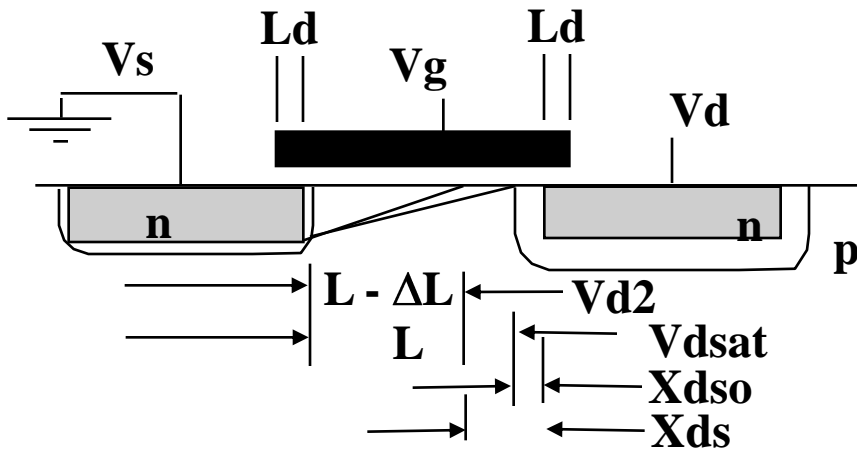


2ND GENERATION EQUATIONS FOR CHANNEL LENGTH MODULATION

KAPPA is channel length modulation parameter.

$$\text{KAPPA is calculated} = [(qN_{\text{sub}}/(2\epsilon_0\epsilon_r))((1-I_{\text{dsat}}/I_{\text{d}'}) (L-2L_{\text{D}}-X_{\text{dso}}-X_{\text{ds}}))^2 / (V_{\text{d2}}-V_{\text{dsat}})]^{0.5}$$

Measure $I_{\text{d}'}$ at large V_{ds} , and I_{dsat} at V_{dsat} ,
Kappa has units of 1/V typical value ~0.1



PARAMETERS FOR SPICE LEVEL 3

SPICE LEVEL 3 MODEL PARAMETERS FOR MOS TRANSISTORS:

Control	Level=3	
Process	TPG=1	1 if gate is doped opposite of channel, -1 if not
Process	TOX	Gate Oxide Thickness
Process	NSUB	Channel doping concentration
Process	XJ	Drain/Source Junction Depth
Process	PB	PB is the junction built in voltage
W and L	LD	Drain/Source Lateral Diffusion
W and L	WD	Decrease in Width from Drawn Value
DC	UO	Zero Bias Low Field Mobility
DC	VTO	Measured threshold voltage, long wide devices
DC	THETA	Gate Field Induced Mobility Reduction
DC	DELTA	Narrow Channel Effect on the Threshold Voltage
DC	VMAX	Maximum Carrier Velocity
DC	ETA	DIBL Coefficient
DC	KAPPA	Channel Length Modulation Effect on Ids
DC	NFS	Surface State Density

PARAMETERS FOR SPICE LEVEL 3

Diode & Resistor RS

Diode & Resistor RD

AC CGDO

AC CGSO

AC CGBO

AC CJ

Temp

- more

Noise

- more

Tunneling

-more

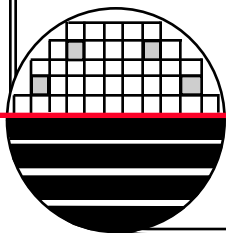
Source Series Resistance

Drain Series Resistance

Zero Bias Gate-Source Capacitance

Zero Bias Gate-Drain Capacitance

Zero Bias Gate-Substrate Capacitance



BSIM3 MODELS

Berkeley SPICE third generation SPICE models are called BSIM3. These models for transistors use equations that are continuous over the entire range of operation (sub-threshold, linear region and saturation region). The equations for mobility are improved. Equations for temperature variation, stress effects, noise, tunneling have been added and/or improved. BSIM3 is presently the industry standard among all these models. It represents a MOSFET with many electrical and structural parameters, among which, only W and L are under the control of a circuit designer. All the rest are fixed for all MOSFETs integrated in a given fabrication technology, and are provided to the designer as an “untouchable” deck of device parameters. (There are over 200 parameters in some versions of BSIM3 models)

SPICE LEVEL-49 EQUATIONS FOR VT

$$\begin{aligned}
 V_{th} = & VTH0 + K1 \cdot (\sqrt{\phi_s} - V_{bseff} - \sqrt{\phi_s}) - K2 \cdot V_{bseff} \\
 & + K1 \cdot \left(\sqrt{1 + \frac{NLX}{L_{eff}}} - 1 \right) \cdot \sqrt{\phi_s} + (K3 + K3_b \cdot V_{bseff}) \cdot \frac{TOX}{W_{eff} + W_0} \cdot \phi_s \\
 & - DVT0 \cdot \left(\exp\left(-DVT1 \cdot \frac{L_{eff}}{2l_t}\right) + 2 \cdot \exp\left(-DVT1 \cdot \frac{L_{eff}}{l_t}\right) \right) \cdot (V_{bi} - \phi_s) \\
 & - \left(\exp\left(-DSUB \cdot \frac{L_{eff}}{2l_{t0}}\right) + 2 \cdot \exp\left(-DSUB \cdot \frac{L_{eff}}{l_{t0}}\right) \right) \cdot (ETA0 + ETAB \cdot V_{bseff}) \cdot V_{ds} \\
 & - DVT0W \cdot \left(\exp\left(-DVT1W \cdot \frac{L_{eff} \cdot W_{eff}}{2l_{tw}}\right) + 2 \cdot \exp\left(-DVT1W \cdot \frac{L_{eff} \cdot W_{eff}}{l_{tw}}\right) \right) \cdot (V_{bi} - \phi_s)
 \end{aligned}$$

Note: Vth0 is from Level=1 equation

SPICE LEVEL-49 EQUATIONS FOR UO

MOBMOD = 1

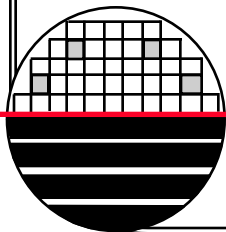
$$\mu_{\text{eff}} = \frac{UO}{1 + (UA + UC \cdot V_{\text{bseff}}) \cdot \left(\frac{V_{\text{gsteff}} + 2 \cdot V_{\text{th}}}{\text{TOX}} \right) + UB \cdot \left(\frac{V_{\text{gsteff}} + 2 \cdot V_{\text{th}}}{\text{TOX}} \right)^2}$$

UA, UB and UC are empirically fit and replace THETA and VMAX used in LEVEL 3

Effective Vgs - Vth

$$V_{\text{gsteff}} = \frac{2 \cdot n \cdot V_t \cdot \ln \left[1 + \exp \left(\frac{V_{\text{gs}} - V_{\text{th}}}{2 \cdot n \cdot V_t} \right) \right]}{1 + 2 \cdot n \cdot \text{COX} \cdot \sqrt{\frac{2 \cdot \phi_s}{q \cdot \epsilon_{\text{si}} \cdot \text{NCH}}} \cdot \exp \left(-\frac{V_{\text{gs}} - V_{\text{th}} - 2 \cdot \text{VOFF}}{2 \cdot n \cdot V_t} \right)}$$

$$n = 1 + \text{NFACTOR} \cdot \text{Cd} / \text{COX} + ((\text{CDSC} + \text{CDSCD} \cdot V_{\text{ds}} + \text{CDSCB} \cdot V_{\text{bseff}}) - (\exp(-\text{DVT1} \cdot \text{Leff} / 2l_t) + 2 \exp(-\text{DVT1} \cdot \text{Leff} / l_t))) / \text{COX} + \text{CIT} / \text{COX}$$



SPICE LEVEL-49 EQUATIONS FOR ID

Effective Vds

$$V_{dseff} = V_{dsat} - \frac{1}{2} \cdot (V_{dsat} - V_{ds} - \text{DELTA} + \sqrt{(V_{dsat} - V_{ds} - \text{DELTA})^2 + 4 \cdot \text{DELTA} \cdot V_{dsat}})$$

Drain Current

$$I_{ds} = \frac{I_{dso}}{1 + \frac{R_{ds} \cdot I_{dso}}{V_{dseff}}} \cdot \left(1 + \frac{V_{ds} - V_{dseff}}{V_A}\right) \cdot \left(1 + \frac{V_{ds} - V_{dseff}}{V_{ASCBE}}\right)$$

$$I_{dso} = \frac{W_{eff} \cdot \mu_{eff} \cdot \text{COX} \cdot V_{gateff} \cdot \left(1 - A_{bulk} \cdot \frac{V_{dseff}}{2 \cdot (V_{gateff} + 2 \cdot V_t)}\right) \cdot V_{dseff}}{L_{eff} \cdot [1 + V_{dseff} / (E_{sat} \cdot L_{eff})]}$$

$$V_A = V_{Asat} \cdot \left(1 + \frac{PVAG \cdot V_{gateff}}{E_{sat} \cdot L_{eff}}\right) \cdot \left(\frac{1}{V_{ACLM}} + \frac{1}{V_{ADIBLC}}\right)^{-1}$$

$$V_{ACLM} = \frac{A_{bulk} \cdot E_{sat} \cdot L_{eff} + V_{gateff}}{PCLM \cdot A_{bulk} \cdot E_{sat} \cdot l_{itl}} \cdot (V_{ds} - V_{dseff})$$

UTMOST III Modeling Manual-Vol.1.
Ch. 5. from Silvaco International.

SPICE LEVEL-49 EQUATIONS FOR ID (cont)

$$V_{ADIBLC} = \frac{(V_{gsoff} + 2 \cdot V_t)}{\theta_{reut} \cdot (1 + PDIBLCB \cdot V_{bsoff})} \cdot \left(1 - \frac{A_{bulk} \cdot V_{dsat}}{A_{bulk} \cdot V_{dsat} + V_{gsoff} + 2 \cdot V_t} \right)$$

$$\theta_{reut} = PDIBLC1 \cdot \left[\exp\left(-DROUT \cdot \frac{L_{eff}}{2 \cdot l_{to}}\right) + 2 \cdot \exp\left(-DROUT \cdot \frac{L_{eff}}{l_{to}}\right) \right] + PDIBLC2$$

$$\frac{1}{V_{ASCBE}} = \frac{PSCBE2}{L_{eff}} \cdot \exp\left(\frac{-PSCBE1 \cdot l_{fil}}{V_{ds} - V_{dsoff}}\right)$$

$$V_{Asat} = \frac{E_{sat} \cdot L_{eff} + V_{dsat} + 2 \cdot R_{ds} \cdot VSAT \cdot COX \cdot W_{eff} \cdot V_{gsoff} \cdot \left[1 - \frac{A_{bulk} \cdot V_{dsat}}{2 \cdot (V_{gsoff} + 2 \cdot V_t)} \right]}{2/\lambda - 1 + R_{ds} \cdot VSAT \cdot COX \cdot W_{eff} \cdot A_{bulk}}$$

$$l_{fil} = \sqrt{\frac{\epsilon_{si} \cdot TOX \cdot XJ}{\epsilon_{ox}}}$$

PARAMETERS FOR SPICE BSIM3 LEVEL 49**SPICE BSIM3 LEVEL 49 MODEL PARAMETERS FOR MOS TRANSISTORS:**

Control	LEVEL=49	
Control	MOBMOD=1	Mobility model selector choice
Control	CAPMOD=1	Capacitor model selector choice
Process	TOX	Gate Oxide Thickness
Process	XJ	Drain/Source Junction Depth
Process	NCH	Channel Surface doping concentration
Process	NSUB	Channel doping concentration
Process	XT	Distance into the well where NCH is valid
Process	NSF	Fast Surface State Density
Process	NGATE	Gate Doping Concentration
W and L	WINT	Isolation Reduction of Channel Width
W and L	LINT	Source/Drain Underdiffusion of Gate

Note: only some of the few hundred parameters

PARAMETERS FOR SPICE BSIM3 LEVEL 49

DC	VTH0	Threshold voltage, Long, Wide Device, Zero Substrate Bias = VTO in level 3
DC	U0	Low Field Mobility, U0 in level 3
DC	PCLM	Channel Length Modulation Parameter
Diode & Resistor	RSH	Drain/Source sheet Resistance
Diode & Resistor	JS	Bottom junction saturation current per unit area
Diode & Resistor	JSW	Side wall junction saturation current per unit length
Diode & Resistor	CJ	Bottom Junction Capacitance per unit area at zero bias
Diode & Resistor	MJ	Bottom Junction Capacitance Grading Coefficient
Diode & Resistor	PB	PB is the junction built in voltage
Diode & Resistor	CJSW	Side Wall Junction Capacitance per meter of length
Diode & Resistor	MJSW	Side Wall Junction Capacitance Grading Coefficient
AC	CGSO	Zero Bias Gate-Source Capacitance per meter of gate W
AC	CGDO	Zero Bias Gate-Drain Capacitance per meter of gate W
AC	CGBO	Zero Bias Gate-Substrate Capacitance per meter of gate L

Note: only some of the few hundred parameters

EXCEL SPREADSHEET SPICE PARAMETER CALCULATOR

	A	B	C	D	E	F	G	H	I	J	K	L
1	ROCHESTER INSTITUTE OF TECHNOLOGY								SPICE_Parameter_Calculator.xls			
2	MICROELECTRONIC ENGINEERING								1/18/2007			
3												
4	CALCULATION OF MOSFET SPICE PARAMETERS								DR. LYNN FULLER			
5												
6	To use this spreadsheet change the values in the white boxes. The rest of the sheet is protected and should not be changed unless											
7	you are sure of the consequences. The calculated results are shown in the purple boxes.											
8												
9	CONSTANTS											
10												
11	T=	300	K		BoronD0	0.76	cm ² /s					
12	KT/q=	0.026	volts		BoronEa	3.46	eV					
13	ni =	1.45E+10	cm ⁻³		PhosphorousD0	3.85	cm ² /s					
14	Eo =	8.85E-14	F/cm		PhosphorousEa	3.66	eV					
15	Er si =	11.7										
16	Er SiO2 =	3.9			Carrier V elocity Saturation occurs at ~ 5E6 to 2E7 cm/s, extracted values can be artificially 2 times higher							
17	E affinity =	4.15	volts		Critical value of electric field Ec of ~8E3 to 3E4 V/cm for electrons, ~2E4 to 1E5 V/cm for holes							
18	q =	1.60E-19	coul									
19	Eg =	1.124	volts									
20												
21	INTRODUCTION											
22												
23	This spreadsheet calculates nmos or pmos level one, three and BSIM3 SPICE parameters from details known about the process parameters, device layout and											
24	fabrication history. Level one spice parameters assume mobility is a function of total impurity concentration and temperature only. Level one uses the parameter											
25	LAMBDA for channel length modulation. Different equations are used to calculate Ids in the saturation and non-saturation regions of operation.											
26												
27	The level three SPICE model is derived from the level one model with some additional parameters to better account for the decrease in carrier mobility for											
28	high vertical and lateral electric fields. The level three model also allows the user to account for narrow channel effects, drain induced barrier lowering (DIBL),											
29	and gives better sub threshold characteristics. For example the parameter LAMBDA is replaced by a more complex model using the parameter VMAX and											
30	KAPPA. The low field mobility value UO is modified for high gate electric fields with parameter THETA and modified for high lateral electric fields through											
31	the VMAX parameter. Different equations are used to calculate Ids in the saturation, non-saturation and subthreshold regions of operation.											
32												
33	The BSM3 SPICE parameters are derived from the level one and three parameters. BSIM models have hundreds of parameters used to fully describe DC and AC											
34	device operation, temperature effects, noise, stress effects and more. Most of the parameters can only be determined from measured device performance. In											
35	this spreadsheet the BSIM3 parameters are derived from level one and level three parameters. All other parameters are not specified and the default values are											
36	invoked. The single equation for Ids is used that is valid in saturation, non-saturation and subthreshold regions of operation, making convergence during circuit											
37	simulation more reliable.											
38												
39												
40	References:	<u>MOSFET Modeling with SPICE</u> , Daniel F oty, 1997, Prentice Hall, ISBN-0-13-227935-5										
41		<u>Operation and Modeling of the MOS Transistor</u> , 2nd Edition, Y annis Tsividis, 1999, McGraw-Hill, ISBN-0-07-065523-5										
42		<u>UTMOST III Modeling Manual-V ol.1</u> , Ch. 5. From Silvaco International.										
43												

[SPICE Parameter Calculator.xls](#)

INPUTS AND RESULTS

	A	B	C	D	E	F	G	H	I	J	K	L
44	LAY OUT PARAMETERS				VALUES CALCULATED FROM PROCESS PARAMETERS							
45	(assume source and drain are symmetrical)											
46	L			2	um			Diffusion Constant at Temp of Well Drive			1.43E-13	cm ² /s
47	W			16	um			Starting wafer doping = $1/(q \text{ um ax Rho})$			4.42E+14	cm ⁻³
48	Area of Drain/Source			96	um ²			Well Surface Concentration= $N_s = \text{Dose} / (\pi D \theta^{0.5})$			1.45E+17	cm ⁻³
49	Perimeter of Drain/Source			44	um			Well Depth = $((4D \text{ dT d/Dose}) \ln(N_{\text{sub}}(\pi D \text{ dT } \theta^{0.5}))^{0.5})$			3.75	um
50	# squares between Contact and Channel			0.143				Well average doping, $N_{\text{ave}} = \text{Dose}/x_j$			5.33E+16	cm ⁻³
51	# squares between LDD/N+ and Channel			0.025				Bulk Well Majority Carrier Mobility at N=N _{ave}			394.32	cm ² /v-s
52								Bulk Well Minority Carrier Mobility at N=N _{ave}			1004.52	cm ² /v-s
53	PROCESS PARAMETERS			1=yes, 0=No								
54	Aluminum gate			0				Well Sheet Resistance = $1/(q(\mu(N_{\text{ave}})D_{\text{ose}}))$			792.50	ohm-s
55	n+ Poly gate			1				Well surface mobility at Surface Doping Concentration			725.76	cm ² /V-s
56	p+ Poly gate			0				$W_{\text{dmax}} = (4 \text{ eosi } \phi_s / q / N_{\text{ave}})^{0.5}$			0.143	um
57	N well (pMOSFET)			0				Metal Work Function, ϕ_m			4.12	volts
58	P well (nMOSFET)			1				Magnitude of Semiconductor Potential (Fermi - Intrinsic), ϕ_s			0.419	volts
59	V t adjust Dose (+ for Boron, - for Phos)			0.00E+00	cm-2			Oxide Capacitance/cm ² = $\text{Cox} = \epsilon_0 \epsilon_{\text{ox}} / d_{\text{ox}}$			2.30E-07	F/cm ²
60	Gate Oxide Thickness			150	Å			Metal Semi Work Function Diff, ϕ_{ms}			-0.170	volts
61	NSS			3.00E+11	cm-2			Flat Band Voltage, V_{FB}			-0.379	volts
62	Starting Wafer Resistivity			10	ohm-cm			Threshold Voltage, V_{TO}			1.33	volts
63	Well Dose			2.00E+13	cm-2			Threshold Adjust, $\Delta V = q D_{\text{ose}} / \text{Cox}$			0.00	volts
64	Well Drive Time			710	min			Ion Implanted Adjusted Threshold Voltage, V_{T}			1.33	volts
65	Well Drive Temperature			1100	C			Diffusion Constant at Temp of D/S Anneal			1.17E-14	cm ² /s
66	LDD D/S Dose			2.50E+13	cm-2			D/S Junction Depth, X_{J}			0.18	um
67	LDD D/S Drive Time			30	min			D/S average doping, $N_{\text{ave}} = \text{Dose}/x_j$			1.36E+18	cm-3
68	LDD D/S Drive Temperature			1000	C			Bulk Mobility in D/S at N=N _{ave}			230.94	cm ² /v-s
69	Field Oxide Thickness			6000	Å			D/S Sheet Resistance = $1/(q(\mu(N_{\text{ave}})D_{\text{ose}}))$			1082.55	ohm-s
70	Minority Carrier Lifetime in the well			1	µs			Lateral Diffusion = $LD = 0.8 * X_{\text{J}}$			0.15	µm
71	D/S Dose (N+ or P+)			2.00E+15	cm-2			Capacitance/cm ² for Field Oxide			5.75E-09	F/cm ²
72								D/S Width of Space Charge Layer at Zero Bias, X_{ds0}			0.152	µm
73								D/S Width of Space Charge Layer at V _{dd} , Bias X_{ds}			0.380	µm
74								$L_{\text{eff}} = L_{\text{mask}} - 2 * LD - 2 * X_{\text{ds0}}$			1.400	µm
75								Built in Voltage for D/S pn junction			0.95	volts
76								Junction reverse bias current density, J_{S}			3.23E-08	A/m ²
77								Junction Capacitance for D/S at zero bias			6.80E-08	F/cm ²
78	MEASURED TRANSISTOR VALUES			Lambda, Calculated, $((L_{\text{max}}/L_{\text{min}})-1)/V_{\text{dd}}$								
79	V _{dd}			5	volts			CALCULATED SPICE PARAMETERS FROM MEASURED VALUES				
80	Magnitude of IDS at V _{gs} =V _{ds} =V _{dd}			5.41	mAmps	=I _{ds}		U _{eff} mobility to match IDS at V _{gs} =V _{ds} =V _{dd}			286.44	cm ² /v-s
81	Magnitude of IDS at V _{gs} =V _{dd} , V _{ds} =V _{dsat}			5.12	mAmps	=I _{dsat}		LAMBDA measured			0.03	1/volts
82	V _{TO} (+ for nmos and - for pmos)			1.1	volts			V _{TO} measured			1.1	volts
83	I _{sub-min}			0.001	mAmps			$J_{\text{S}} = I_{\text{sub-min}}/\text{Area of Drain measured}$			1.04E-02	A/m ²
84	D/S Sheet Resistance			39.2	ohm-s			RSH measured			39.2	ohm-s
85	Lambda			0.03	1/volts							
86												

PARAMETERS FOR SPICE LEVEL 1

	A	B	C	D	E	F	G	H	I	J	K	L
87	SPICE PARAMETERS FOR LEVEL ONE MODEL											
88	1 The parameters in the yellow boxes are calculated from the other parameters and thus should not be entered in the SPICE model											
89	2 If the SPICE parameters from measured values are different than the calculated SPICE parameters you might want to use them instead.											
90	3 We assume the model definition has L, W, AD, AS, PD, PS, NRS, NRD specified for calculation of some of the parameters in the yellow boxes											
91	4 Lambda is different for every different length transistor in the level one model, so a different model is needed for each different length mosfet											
92												
93	SPICE	Name	SPICE Parameters									
94	Parameter		Using Process Parameters		<i>note: most parameters use 0 not 0 at end of parameter name ("oh" not "zero")</i>							
95	1	Level	1		Schichman and Hodges Model							
96	2	VTO	1.33	volts	Zero Bias Threshold Voltage, enter value if threshold adjust implant is used							
97	3	KP	2.50E-04	F/s-volt	Transconductance Parameter, $KP = UO \ \epsilon_{si} \ \epsilon_o / Tox$							
98	4	GAMMA	9.52E-01	(volts) ^{1/2}	Bulk Threshold Parameter, $GAMMA = [2q \ \epsilon_{si} \ \epsilon_o \ NSUB / C_{ox}]^{1/2}$							
99	5	PHI	0.419	volts	PHI is the semiconductor potential, Intrinsic Level to Fermi Level difference							
100	6	LAMBDA	0.031	1/volts	Channel length modulation parameter							
101	7	RD	27.06	ohms	Series Drain Resistance							
102	8	RS	27.06	ohms	Series Source Resistance							
103	9	CBD	7.08E-14	F	CBD zero bias bulk to drain junction capacitance, $CBD = CJAD + CJSWPD$							
104	10	CBS	7.08E-14	F	CBD zero bias bulk to source junction capacitance, $CBD = CJAD + CJSWPD$							
105	11	IS	3.10E-18	A	D/S junction leakage current							
106	12	PB	0.95	volts	PB is the junction built in voltage, $PB = (KT/q) \ln(NSUB/n_i) + 0.56$							
107	13	CGSO	3.40E-10	F/m	G-to-S overlap C (per m channel width) $CGSO = Cox * (mask \ overlap \ in \ L \ direction + LD)$							
108	14	CGDO	3.40E-10	F/m	G-to-D overlap C (per m channel width) $CGDO = Cox * (mask \ overlap \ in \ L \ direction + LD)$							
109	15	CGBO	5.75E-10	F/m	G-to-well overlap C (per meter channel length) $CGSO = Cox * (mask \ overlap \ in \ W \ direction)$							
110	16	RSH	1082.55	ohms	Sheet resistance of D/S							
111	17	CJ	6.80E-04	F/m ²	D/S Bottom junction capacitance/m ² ,							
112	18	MJ	0.5		Junction Grading Coefficient for bottom of D/S Junction							
113	19	CJSW	1.26E-10	F/m	D/S side wall junction capacitance per meter of D/S perimeter							
114	20	MJSW	0.5		Junction Grading Coefficient for side of D/S Junction							
115	21	JS	3.23E-08	A/m ²	D/S junction leakage current							
116	22	TOX	1.50E-08	m	Gate Oxide Thickness							
117	23	NSUB	1.45E+17	cm-3	Well Doping Nave							
118	24	NSS	3.00E+11	cm-2	Surface State Density as known from process knowledge							
119	25	NFS	0		Fast Surface States, Always set to zero							
120	26	TPG	1		+1 if gate doped opposite of channel, -1 if gate doped same as channel, 0 if gate is aluminum							
121	27	XJ	0.18	um	D/S Junction Depth							
122	28	LD	0.15	um	Lateral Diffusion of D/S into the channel arbitrarily set to 80% of XJ							
123	29	UO	363	cm ² /v-s	Well surface minority carrier mobility at well surface concentration <i>divided by two</i>							
124	MODEL RITSUBN1 NMOS (LEVEL=1						MODEL RITSUBP1 PMOS (LEVEL=1					
125	+VTO=1.0 LAMBDA=0.031 PB=0.95 CGSO=3.4E-10 CGDO=3.4E-10						+VTO=1.0 LAMBDA=0.05 PB=0.94 CGSO=5.08E-10 CGDO=5.08E-10					
126	+CGBO=5.75E-10 RSH=1082 CJ=6.8e-4 MJ=0.5 CJSW=1.26e-10						+CGBO=5.75E-10 RSH=33.7 CJ=5.01e-4 MJ=0.5 CJSW=1.38e-10					
127	+MJSW=0.5 JS=3.23e-8 TOX=150E-10 NSUB=1.45e17 NSS=3E11						+MJSW=0.5 JS=6.43e-8 TOX=150E-10 NSUB=7.23e16 NSS=1E11					
128	+TPG=+1 XJ=0.18U LD=0.15U UO=363)						+TPG=+1 XJ=0.28U LD=0.22U UO=363)					

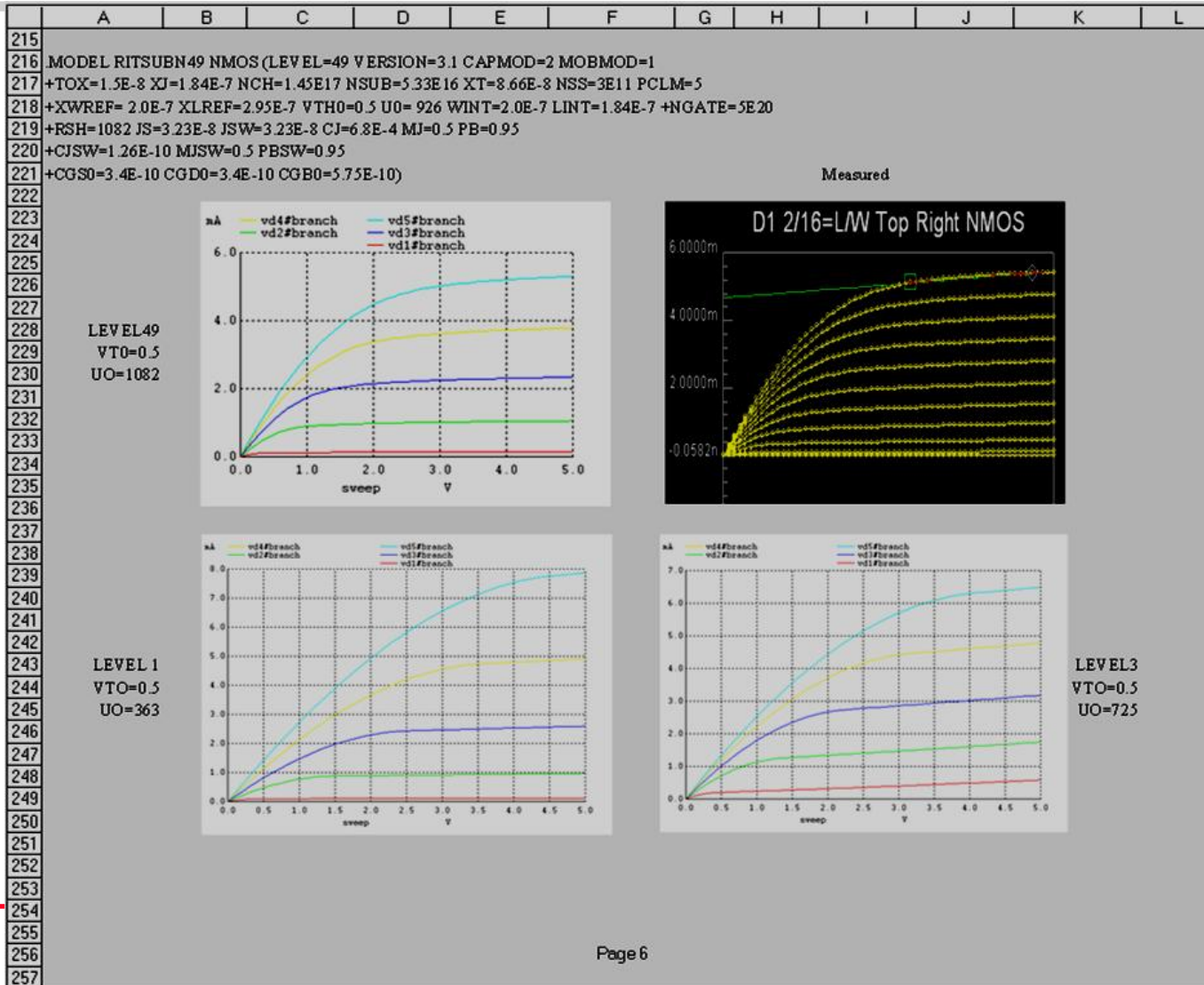
PARAMETERS FOR SPICE LEVEL 3

	A	B	C	D	E	F	G	H	I	J	K	L
129	SPICE PARAMETERS FOR LEVEL THREE MODEL											
130	1 WD is estimated to be 1/2 the field oxide thickness for a LOCOS process											
131	2 THETA is calculated from $U_{eff} = U_0 / (1 + THETA(V_{gs} - V_t))$ and $I_{ds} = U_{eff} (C_{ox}/L) (W/L) (V_{gs} - V_t)^2 (1 + \lambda V_{ds})$ using measured I_{ds} and V_t values											
132	3 DELTA is calculated = $q * N_{ave} * (X_{ds})^2 / (s_0 \cdot s_i (2 \cdot \phi_s))$											
133	4 KAPPA is calculated = $[(q N_{sub} / (2 \cdot s_0 \cdot \alpha))] ((1 - I/T) (L - 2LD - X_{dso} - X_{ds}))^2 / (V_{ds} - V_{dsat})^{0.5}$											
134	5 VMAX is calculated from effective mobility times electric field at $V_{gs} = V_{ds} = V_{dsat}$, where $E = V_{dsat} / L_{eff}$											
135	6 ETA is calculated from the ratio of charge in the channel at $V_{ds} = V_{dd}$ to charge in the channel at $V_{ds} = 0$											
136	note: Parameters in Red come directly from SPICE Level One											
137	Parameter	Name	Value	Units	note: most parameters use 0 not 0 at end of parameter name ("oh" not "zero")							
138	1	Level	3									
139	2	TPG	1		Type of Gate Material							
140	3	TOX	1.50E-08	m	Gate Oxide Thickness							
141	4	LD	2.95E-07	m	Channel Length Reduction from Drawn Value							
142	5	WD	3.00E-07	m	Channel Width Reduction From Drawn Value							
143	6	UO	726	cm ² /V-s	Zero Bias Low Field Mobility							
144	7	VTO	1.33	V	Measured threshold voltage for long wide devices with zero substrate bias							
145	8	THETA	0.393	1/V	Gate Field Induced Mobility Reduction Parameter							
146	9	RS	27.06	ohm	In level 3 only lumped resistance is available, each different width FET has a different model							
147	10	RD	27.06	ohm	In level 3 only lumped resistance is available, each different width FET has a different model							
148	11	DELTA	2.27		Narrow Channel Effect on the Threshold Voltage							
149	12	NSUB	1.45E+17	cm-3	Effective Substrate Doping							
150	13	XJ	1.84E-07	m	Drain/Source junction depth							
151	14	VMAX	1.02E+07	m/s	Maximum Carrier Velocity (extraction can give 1.2 to 2 times expected saturation velocity)							
152	15	ETA	0.837		DIBL Coefficient							
153	16	KAPPA	0.509	1/V	Channel Length Modulation Effect on the Drain Current							
154	17	NFS	3.00E+11	cm-2	Surface State Density							
155	18	CGSO	3.40E-10	F/m	Zero Bias Gate-Source Capacitance							
156	19	CGDO	3.40E-10	F/m	Zero Bias Gate-Drain Capacitance							
157	20	CGBO	5.75E-10	F/m	Zero Bias Gate-Substrate Capacitance							
158	21	PB	0.95	V	PB is the junction built in voltage, $PB = (KT/q) \ln(NSUB/ni) + 0.56$							
159	22	XQC	0.40		Charge Partitioning Parameter (from Ward and Dutton)							
160	A different model is needed for each transistor of different length or width. Example models shown below.											
161												
162	*.MODEL RITSUBN3 NMOS (LEVEL=3 TPG=1 TOX=1.5E-8 LD=2.95E-7 WD=3.00E-7											
163	*+U0= 726 VTO=0.5 THETA=0.393 RS=27 RD=27 DELTA=2.27 NSUB=1.45E17											
164	*+XJ=1.84E-7 VMAX=1.10E7 ETA=0.837 KAPPA=0.509 NFS=3E11											
165	*+CGSO=3.4E-10 CGDO=3.48E-10 CGBO=5.75E-10 PB=0.95 XQC=0.4)											
166												
167	*.MODEL RITSUBP3 PMOS (LEVEL=3 TPG=1 TOX=1.5E-8 LD=3.61E-7 WD=3E-7											
168	+U0=377 VTO=-0.93 THETA=0.32 RS=33.7 RD=33.7 DELTA=2.35 NSUB=7.12E16											
169	+XJ=2.26E-7 VMAX=3.84E6 ETA=0.897 KAPPA=4.481 NFS=3E11											
170	+CGSO=4.15E-10 CGDO=4.15E-10 CGBO=5.75E-10 PB=0.94 XQC=0.40)											
171												

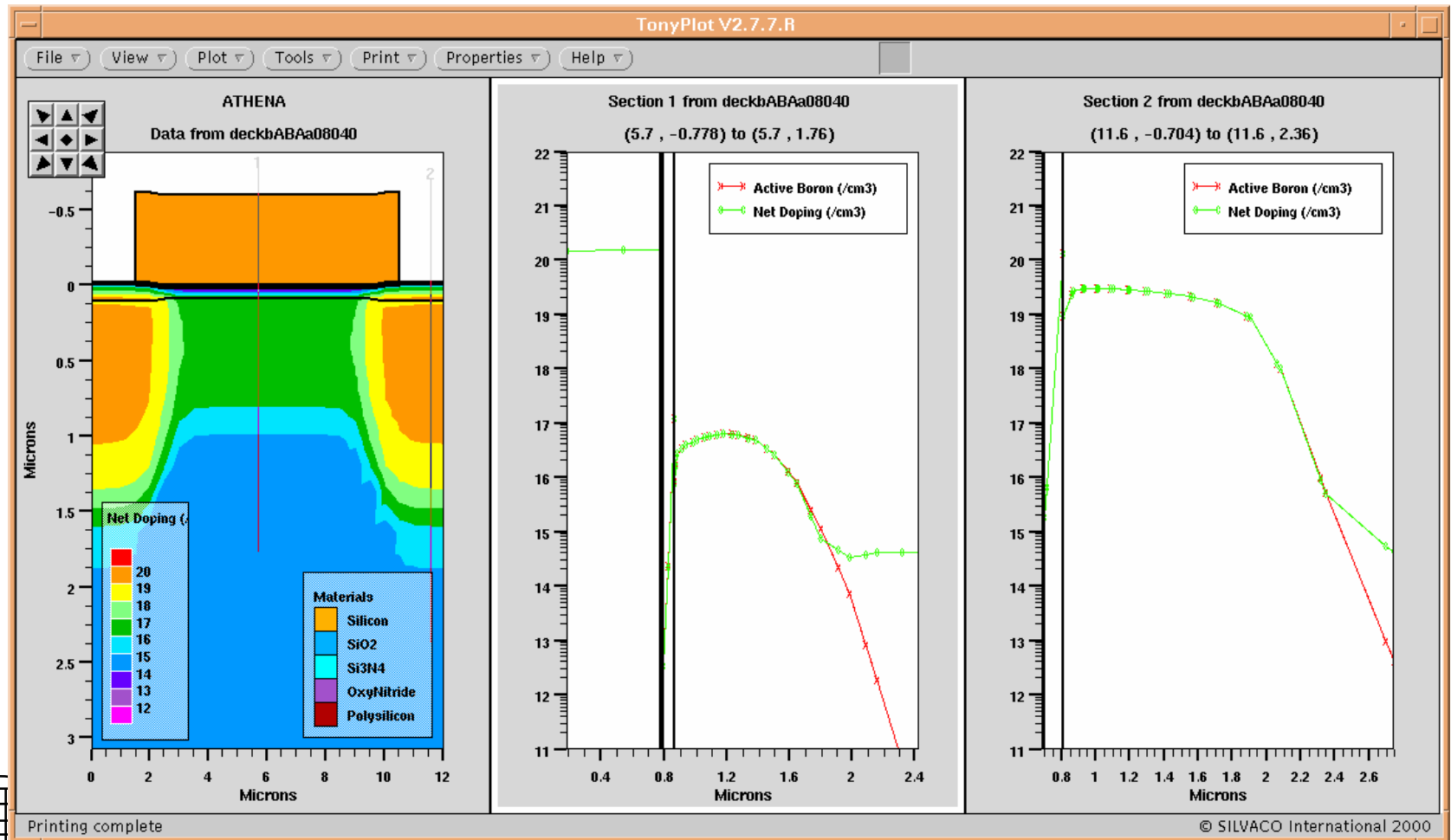
PARAMETERS FOR SPICE LEVEL 49

	A	B	C	D	E	F	G	H	I	J	K	L
172	SPICE PARAMETERS FOR BISIM3 VER 3.1, LEVEL 49											
173	BSIM3V3 is the industry standard, physics-based, deep submicron MOSFET SPICE model for digital and analog circuit design from the Device Group at the											
174	University of California at Berkeley. Level 8 is the original Berkeley version, Level 81 is a slightly modified Silvaco version, Level 49 and 53 are Hspice versions.											
175												
176	note: most parameters use 0 not O at end of parameter name ("zero" not "oh")											
177	note: Parameters in Red come directly from SPICE Level One and/or Three											
178	Parameter	Name	Value	Units								
179	Control	Level	49		Level 8, 81, 49 or 53							
180	Control	VERSION	3.1		3.0, 3.1 or 3.2 versions, default is the newest version							
181	Control	MOBMOD	1		Mobility model selector (1,2,3,4... selects slightly different equations for calculation of U _{eff})							
182	Control	CAPMOD	2		Capacitance model selector (1,2,3,4... selects slightly different equations for gate C _{eff})							
183	Process	TOX	1.50E-08	m	Gate oxide thickness							
184	Process	XJ	1.84E-07	m	Junction Depth							
185	Process	NCH	1.45E+17	cm-3	Well surface doping concentration							
186	Process	NSUB	5.33E+16	cm-3	Well doping concentration below the surface							
187	Process	XT	1.43E-07	m	Distance into well where surface concentration is valid, Default = 1.5E-7m							
188	Process	NSS	3.00E+11	cm-2	Surface State Density, Level 3 NFS or Level 1 NSS treated as equal							
189	W and L	XWREF	2.0E-07	m	Isolation Reduction of Channel Width (from process knowledge)							
190	W and L	XLREF	2.95E-07	m	Source/Drain Underdiffusion of Gate							
191	DC	VTH0	1.33	V	Threshold voltage, Long Wide Device, Zero Substrate Bias = V _{TO} in level 3							
192	DC	U0	725.76	cm ² /v-s	Low Field Mobility							
193	DC	WINT	2.0E-07	m	Isolation Reduction of Channel Width (from process knowledge)							
194	DC	LINT	1.84E-07	m	Source/Drain Underdiffusion of Gate (set equal to XJ)							
195	DC	PCLM	5.00		Channel Length Modulation Parameter, default = 1.3 (select to fit I _{ds} vs. V _{ds} family)							
196	DC	NGATE	5.00E+20	m-3	Gate Doping (SE20 if Diffusion Doped, Dose/Poly Thickness if Ion Implanted with D/S)							
197	Diode/Resistor	RSH	1082.55	ohm/sq	Drain/Source sheet Resistance							
198	Diode/Resistor	JS	3.23E-08	A/m ²	Bottom junction saturation current per unit area							
199	Diode/Resistor	JSW	3.23E-08	A/m	side wall junction saturation current per unit length							
200	Diode/Resistor	CJ	6.80E-04	F/m ²	Bottom Junction Capacitance per unit area at zero bias							
201	Diode/Resistor	MJ	0.5		Bottom Junction Capacitance Grading Coefficient							
202	Diode/Resistor	PB	0.95	V	PB is the junction built in voltage, PB = (KT/q)ln (NSUB/ni) + 0.56							
203	Diode/Resistor	CJSW	1.26E-10	F/m	Side Wall Junction Capacitance per meter of length at zero bias							
204	Diode/Resistor	MJSW	0.5		Side Wall Junction Capacitance Grading Coefficient							
205	Diode/Resistor	PBSW	0.95	V	PBSW is the side wall junction built in voltage, PB = (KT/q)ln (NSUB/ni) + 0.56							
206	AC	CGS0	3.40E-10	F/m	Zero Bias Gate-Source Capacitance per meter of gate width							
207	AC	CGD0	3.40E-10	F/m	Zero Bias Gate-Drain Capacitance per meter of gate width							
208	AC	CGB0	5.75E-10	F/m	Zero Bias Gate-Substrate Capacitance per meter of gate length							
209												
210												
211												
212												
213												
214												

RESULTS USING SPICE LEVELS 49, 3, 1



SILVACO ATHENA SIMULATIONS OF D/S IMPLANT



SILVACO ATHENA (SUPREM)

```

go athena
# set grid
line x loc=0.0 spac=0.1
line x loc=1.0 spac=0.05
line x loc=10.0 spac=0.05
line x loc=12.0 spac=0.1

```

```

line y loc=0.0 spac=0.01
line y loc=2.2 spac=0.01
line y loc=3.5 spac=0.3
line y loc=6.0 spac=0.5

```

```

init silicon phosphor resistivity=11.3 orientation=100 space.mult=5.0

```

```

# ramp up from 800 to 900°c soak 50 min dry o2, ramp down to 800 n2
diff time=10 temp=800 t.final=900 dryo2 press=1.0 hcl.pc=0
diff time=50 temp=900 weto2 press=1.0 hcl.pc=0
diff time=20 temp=900 t.final=800 nitro press=1.0 hcl.pc=0

```

```

deposit photoresist thickness=1.0
etch photoresist left ;1.x=2.0
etch photoresist right p1.x=10.00

```

```

# ion implant drain and source
implant boron dose=1e15 energy=70 tilt=0 rotation=0 crysatal lat.ratio1=1.0 lat.ratio2=1.0

```

```

Etch photoresist all

```

```

# ramp up from 800 to 1000°c soak 90 min, ramp down to 800 n2
diff time=20 temp=800 t.final=1000 nitro press=1.0 hcl.pc=0
diff time=90 temp=1000 nitro press=1.0 hcl.pc=0
diff time=40 temp=1000 t.final=800 nitro press=1.0 hcl.pc=0

```

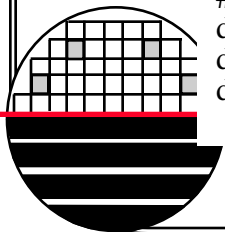
Starting wafer resistivity = 11.3 ohm-cm

Grow Kooi oxide 1000 Å

Ion Implant P-type D/S at Dose = 1E15

Strip photoresist

Anneal D/S implant



SILVACO ATHENA (SUPREM)

Ion Implant P-type channel at
Dose = 0, 4e11, 1e12, 4e12

```
# ion implant channel
implant boron dose=4e12 energy=60 tilt=0 rotation=0 crysatal lat.ratio1=1.0 lat.ratio2=1.0
```

```
etch oxide all
```

```
# ramp up from 800 to 1000°C soak 90 min dry o2, ramp down to 800 n2
diff time=20 temp=800 t.final=1000 dryo2 press=1.0 hcl.pc=0
diff time=90 temp=1000 dryo2 press=1.0 hcl.pc=0
diff time=40 temp=1000 t.final=800 nitro press=1.0 hcl.pc=0
```

Grow 700 Å gate oxide

```
deposit nitride thick=0.010
```

Deposit 100 Å nitride

```
# ramp up from 800 to 1000°C soak 50 min dry o2, ramp down to 800 n2
diff time=10 temp=800 t.final=1000 dryo2 press=1.0 hcl.pc=0
diff time=50 temp=1000 dryo2 press=1.0 hcl.pc=0
diff time=20 temp=1000 t.final=800 nitro press=1.0 hcl.pc=0
```

Temp cycle for growth of oxynitride

```
deposit oxynitride thick=0.01
```

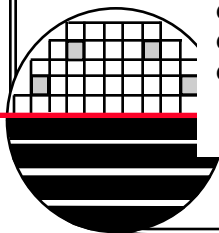
Deposit 100 Å oxynitride

```
deposit poly thick=0.60 c.boron=4e20
```

Deposit 6000 Å poly

```
# ramp up from 800 to 1000°C soak 30 min, ramp down to 800 n2
diff time=20 temp=800 t.final=1000 nitro press=1.0 hcl.pc=0
diff time=30 temp=1000 nitro press=1.0 hcl.pc=0
diff time=40 temp=1000 t.final=800 nitro press=1.0 hcl.pc=0
```

Temp cycle for poly dope



SILVACO ATHENA (SUPREM)

```

etch poly left p1.x=1.5
etch poly right p1.x=10.5

etch oxynitride left p1.x=1.5
etch oxynitride right p1.x=10.5

etch nitride left p1.x=1.5
etch nitride right p1.x=10.5

etch oxide left p1.x=1.5
etch oxide right p1.x=10.5

deposit alumin thick=0.5

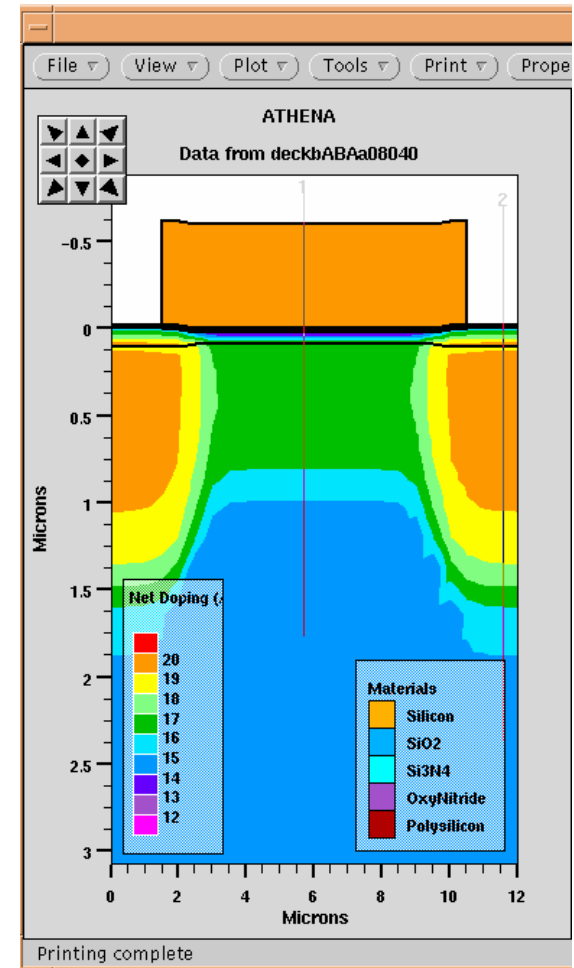
etch alum start x=1.0 y= -2.0
etch cont x=1.0 y= 2.0
etch x=11.0 y= 2.0
etch done x=11.0 y= -2.0

struct outfile=UofH.str

tonyplot UofH.str

quit
    
```

Deposit 5000 Å aluminum

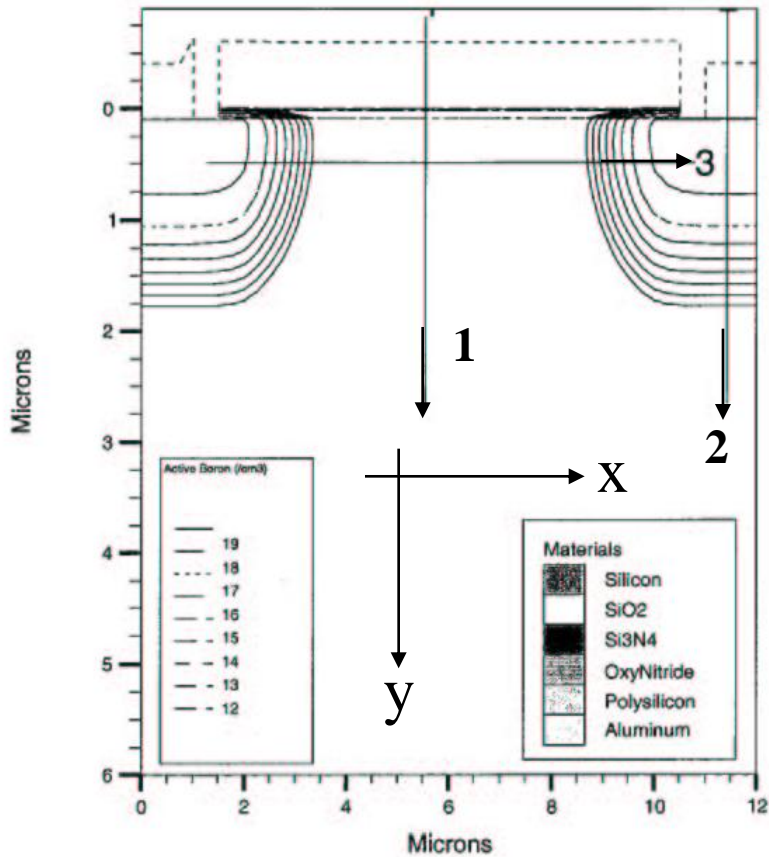


Tonyplot example Only

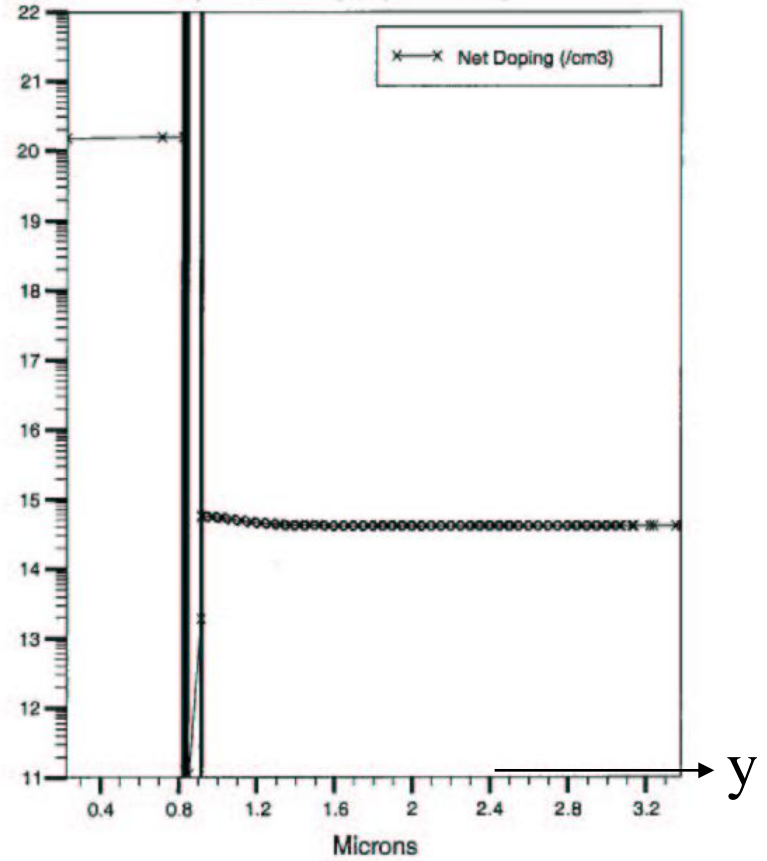
SILVACO ATHENA (SUPREM)

Channel Implant Dose = 0

Crosssection of MOSFET



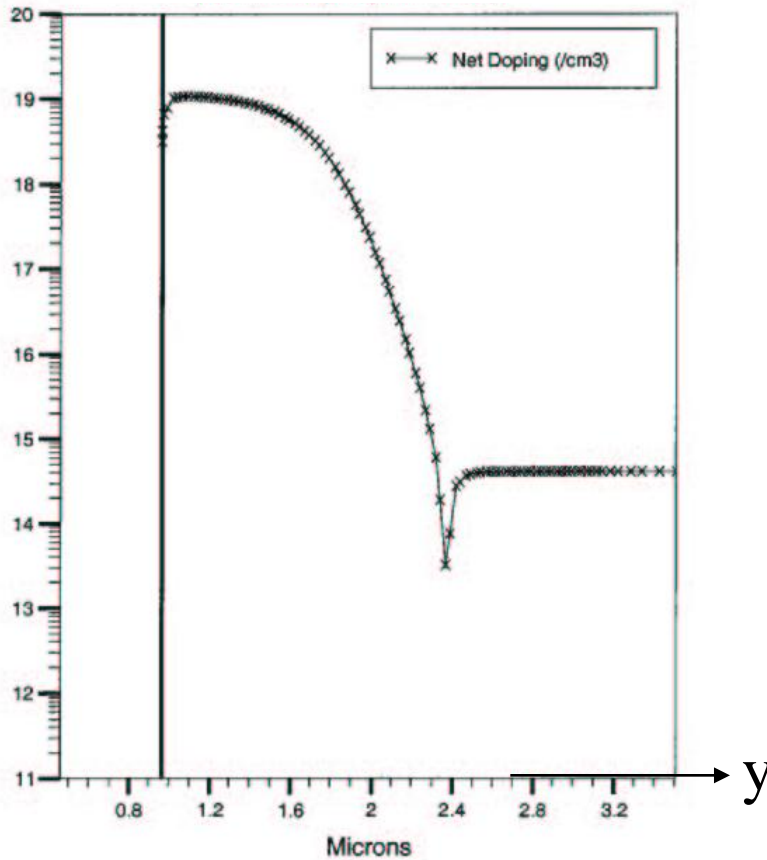
Channel Doping Profile 1



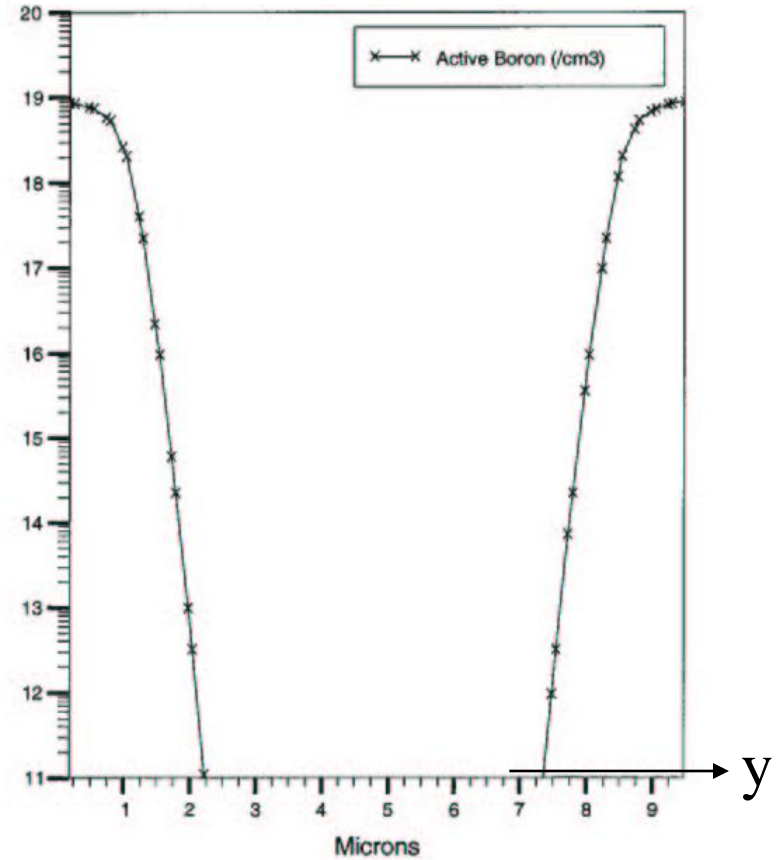
SILVACO ATHENA (SUPREM)

Channel Implant Dose = 0

D/S Doping Profile 2



Channel Doping Profile 3



Microelectronic Engineering

SILVACO ATLAS (DEVICE SIMULATOR)

Go athena
Init infile=UofH.str **Read in structure file created by Athena**

```
#name the electrodes...
Electrode name=gate x=6
Electrode name=source x=0
Electrode name=drain x=12
Electrode name=substrate backside
```

Define location of gate, source, drain and substrate

Extract name="vt" ldvt ptype qss=1e11 workfunc=5.1 x.val=6

Go atlas

```
# define the gate workfunction
Contact name=gate p.poly
# define the Gate qss
Interface qf=1e11
```

```
# use the cvt mobility model for MOS
Models cvt srh
```

```
# set gate biases with Vds=0.0
Solve init
Solve vgate=0 vsubstrate=0 outf=solve_temp0
Solve vgate=-1 vsubstrate=0 outf=solve_temp1
Solve vgate=-1 vsubstrate=0 outf=solve_temp2
Solve vgate=-3 vsubstrate=0 outf=solve_temp3
Solve vgate=-4 vsubstrate=0 outf=solve_temp4
Solve vgate=-5 vsubstrate=0 outf=solve_temp5
```

Do calculations for given gate voltage and substrate voltage ($V_g=0,-1,-2,-3,-4,-5$ and $V_{sub}=0,+5,+10,+15$)

```
# load in temporary file and ramp Vds
Load infile=solve_temp0
Log outf=Vg_0.log
Solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

**Sweep drain voltage from 0 to -5 volts
In -0.5 volt steps**

SILVACO ATLAS (DEVICE SIMULATOR)

```
# load in temporary file and ramp vds
load infile=solve_temp1
log outf=vg_1.log
solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

```
# load in temporary file and ramp vds
load infile=solve_temp2
log outf=vg_2.log
solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

```
# load in temporary file and ramp vds
load infile=solve_temp3
log outf=vg_3.log
solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

```
# load in temporary file and ramp vds
load infile=solve_temp4
log outf=vg_4.log
solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

```
# load in temporary file and ramp vds
load infile=solve_temp5
log outf=vg_5.log
solve name=drain vdrain=0 vfinal=-5 vstep=-0.5
```

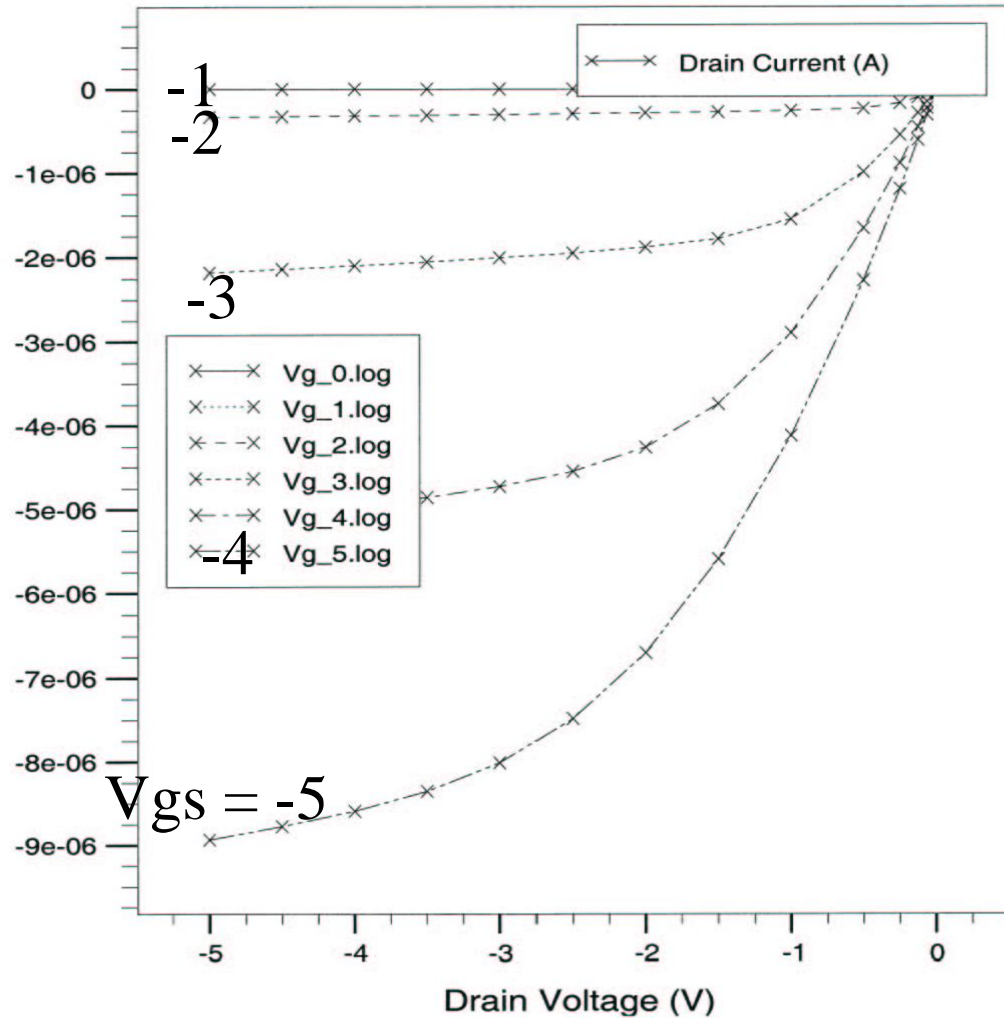
```
# extract max current and saturation slope
extract name="pidsmax" max(abs(i."drain"))
extract name="p_sat_slope" slope(minslope(curve(abs(v."drain"), abs(i."drain"))))
```

```
tonyplot -overlay vg_0.log vg_1.log vg_2.log vg_3.log vg_4.log vg_5.log -setmos1ex09_1.set
quit
```

Sweep drain voltage from 0 to -5 volts
in -0.5 volt steps

ATLAS SIMULATED FAMILY OF CURVES

**Channel Implant
Dose = none
Vsub = 0**



SILVACO ATHENA > ATLAS > UTMOST > SPICE

Extraction of SPICE Model Parameters from ATLAS Device Simulation Using UTMOST

Many users would like to extract SPICE models from their process and device simulation using ATHENA and ATLAS to be used in actual circuit simulation without actually fabricating the device.

Using SILVACO's UTMOST you can extract SPICE model parameters from the simulation results of ATHENA and ATLAS.

To guide users on how to go about extracting SPICE model parameters an example which extracts BSIM3v3 model from process/device simulation is used in this article.

All these commands can be executed from a single software – DeckBuild.

The commands are heavily commented so that you know their functions and purpose.

Here we will concentrate on the UTMOST batch mode commands. Here we only cover a very simple case and there is no local optimization. The UTMOST interactive mode can be used save the UTMOST setup into a file. UTMOST interactive cannot be executed from DeckBuild.

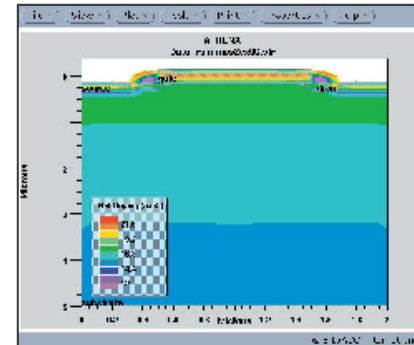


Figure 1. TonyPlot of Device structure.

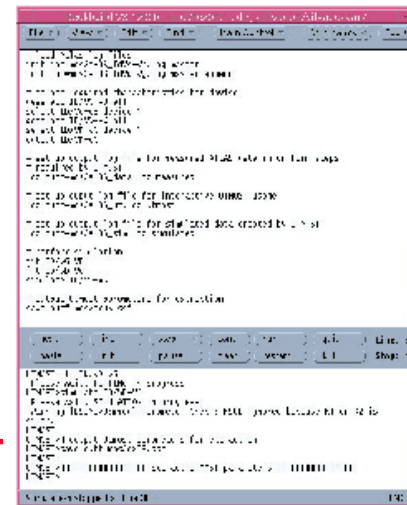


Figure 2. DeckBuild screen.

```

----- start of deckbuild commands -----

# Commands that can be used in deckbuild to extract
# Spice model Parameters. The deck for TCAD is
# not complete. Below are examples of commands that
# maybe used in VWF Athena and Atlas to obtain the
# device characteristics

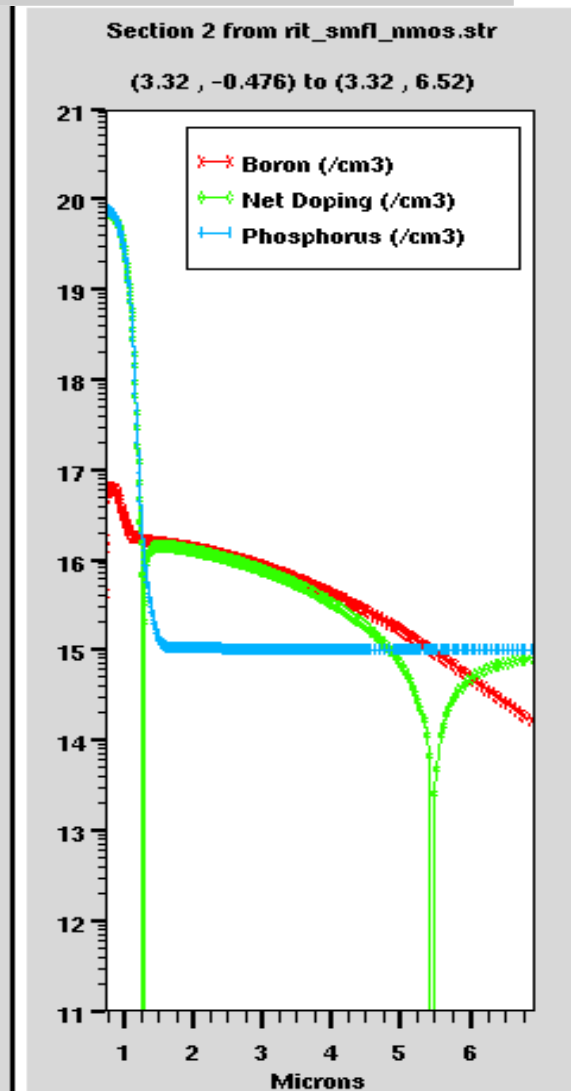
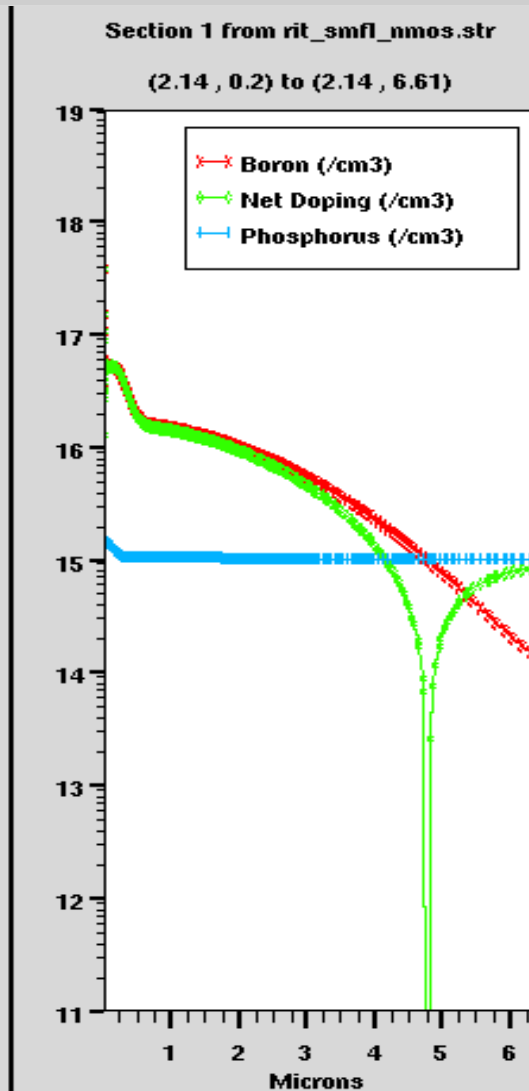
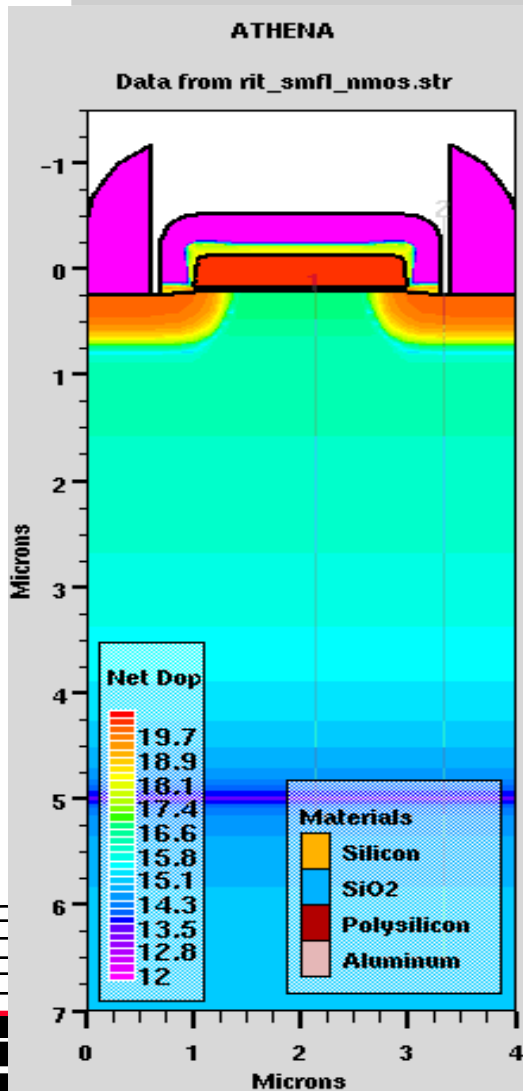
##### START ATHENA SIMULATION ###
# Run process simulation
go athena

# Extract the poly length LD
extract name="ld" thick poly yval=0
extract name="utmmost_ld" ($ld*1.0e-4)
....
extract thickness oxide mat.ocno=1 name="tox"
extract name="utmmost_tox" ($tox*1.0e-10)
....

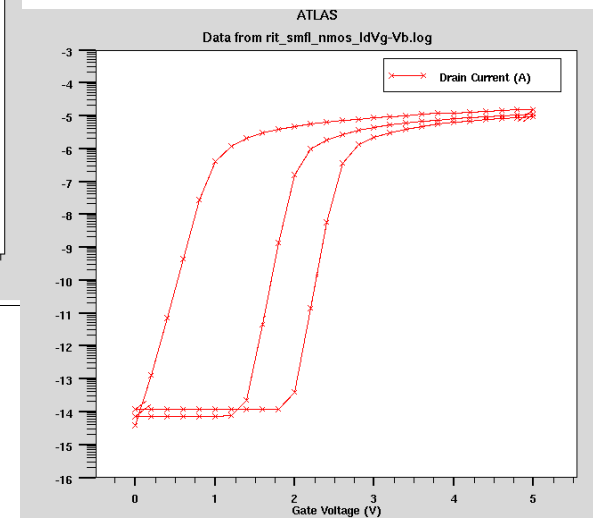
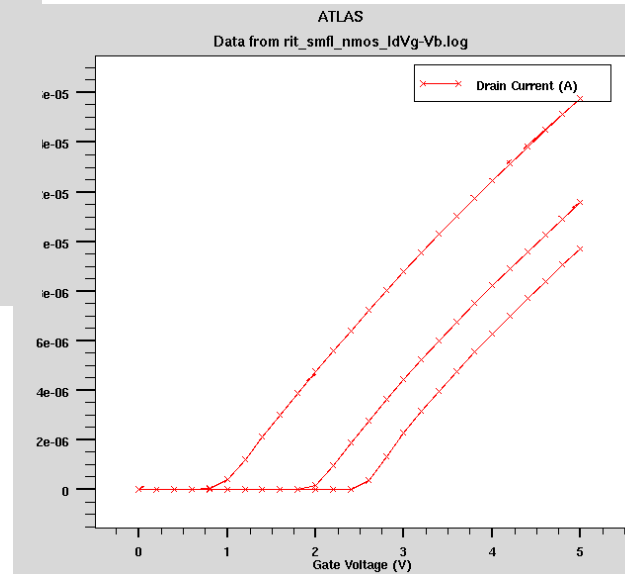
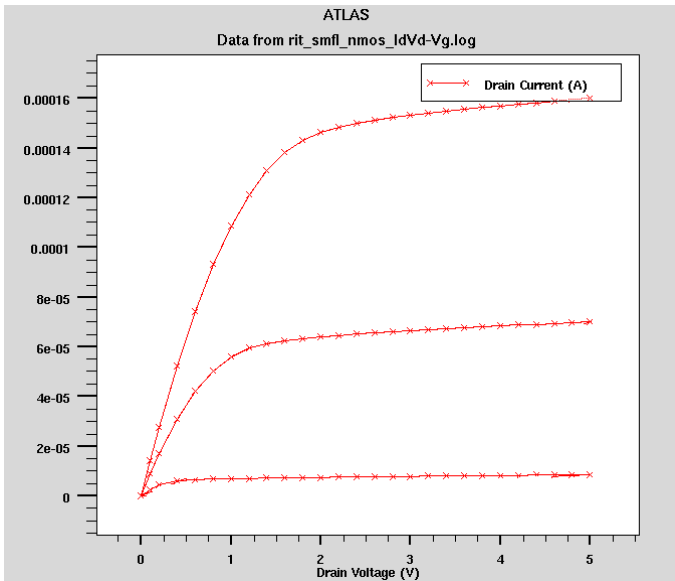
##### START ATLAS DEVICE SIMULATION ##
go atlas
    
```

UTMOST Generates
SPICE model parameters
from ATLAS output file.

SILVACO ATHENA GENERATED IMPURITY PROFILES



ATLAS GENERATED DEVICE CHARACTERISTICS



UTMOST GENERATED SPICE PARAMETERS

NMOS PARAMETER DECK:

```

*2-27-2007 UTMOST EXTRACTIONS.MODEL CMOSN NMOS (LEVEL=49
VERSION=3.1 CAPMOD=2 MOBMOD=1+TOX=328.4E-10 XJ=3.5E-7 NCH=7.0E19
VTH0=0.8627+K1=0.5 K2=-0.0186 K3=80 WO=2.5E-6 NLX=1.740E-7+DVT0W=0
DVT1W=0 DVT2W=-0.032 DVT0=2.2 DVT1=0.53 DVT2=0.1394+U0=670 UA=2.25E-9
UB=5.87E-19 UC=-4.65E-11 VSAT=80000+A0=1 AGS=0 B0=0 B1=0 KETA=-0.047
A1=0 A2=1+RDSW=0 PRWG=0 PRWB=0 WR=1 WINT=2.58E-8 LINT=1.86E-8+XL=0
XW=0 DWG=0 DWB=0 VOFF=-0.06464 NFACTOR=1.3336+CIT=0 CDSC=0.00024
CDSCD=0 CDSCB=0 ETA0=0.08 ETAB=-0.07+DSUB=0.56 PCLM=1.39267
PDIBLC1=0.39 PDIBLC2=0.0086 PDIBLCB=0 +DROUT=0.19093 PSCBE1=4.00E8
PSCBE2=6E-6 PVAG=0 DELTA=0.01 PRT=0+UTE=-1.5 KT1=0 KT1L=0 KT2=0
UA1=4.3E-9 UB1=-7.6E-18+UC1=-5.6E-11 AT=3.3E4 WL=0 WLN=1 WW=0
WWN=1+WWL=0 LL=0 LLN=1 LW=0 LWN=1 LWL=0+XPART=0 +CGD0=1.99E-10
CGS0=1.99E-10 CGB0=5.75E-10 CJ=4.23E-4+PB=0.99 MJ=0.4496 CJSW=3.83
PBSW=0.1083 MJSW=0.1084+PVTH0=0.02128 PRDSW=-16.155 PK2=0.0253
WKETA=0.01886 LKETA=0.0205)**

```

UTMOST GENERATED SPICE PARAMETERS FROM ATHENA SIMULATED DEVICE CHARACTERISTICS

PMOS PARAMETER DECK:

```
*2-27-2007 UTMOST EXTRACTIONS.MODEL CMOSF PMOS (LEVEL=49
VERSION=3.1 CAPMOD=2 MOBMOD=1+TOX=328.7E-10 XJ=3.5E-7 NCH=3.0E19
VTH0=-0.6322+K1=0.6423 K2=-0.0856046 K3=80 K3B=0 WO=2.0E-6 NLX=1.0E-
7+DVT0W=0 DVT1W=0 DVT2W=-0.032 DVT0=1.5 DVT1=0.50 DVT2=-
0.0193+U0=187.362 UA=1.1762E-9 UB=1.0E-22 UC=5.003E-3
VSAT=4.835E6+A0=3.9669 AGS=0 B0=0 B1=0 KETA=-0.0385 A1=0.19469
A2=0.40150+RDSW=0 PRWG=0 PRWB=0 WR=1 WINT=1.67E-8 LINT=3.150E-
7+XL=0 XW=0 DWG=0 DWB=0 VOFF=-0.06464 NFACTOR=1.3336+CIT=0
CDSC=0.00024 CDSCD=0 CDSCB=0 ETA0=0.08 ETAB=-0.07+DSUB=0.56
PCLM=1.39267 PDIBLC1=0 PDIBLC2=1E-5 PDIBLCB=0 +DROUT=0.19093
PSCBE1=4E8 PSCBE2=6E-6 PVAG=0 DELTA=0.01 PRT=0+UTE=-1.5 KT1=0 KT1L=0
KT2=0 UA1=4.3E-9 UB1=-7.6E-18+UC1=-5.6E-11 AT=3.3E4 WL=0 WLN=1 WW=0
WWN=1+WWL=0 LL=0 LLN=1 LW=0 LWN=1 LWL=0+XPART=0 +CGD0=2.4E-10
CGS0=2.4E-10 CGB0=5.75E-10 CJ=7.27E-4+PB=0.97 MJ=0.496 CJSW=3.115
PBSW=0.99 MJSW=0.2654+PVTH0=0.00942 PRDSW=-231.3 PK2=1.397
WKETA=1.863 LKETA=5.729)*
```

UTMOST GENERATED SPICE DECK FROM MEASURED SMFL CMOS PROCESS DEVICE CHARACTERISTICS

*1-15-2007 FROM ROB SAXER UTMOST EXTRACTIONS

```
.MODEL RITSMFLN49 NMOS (LEVEL=49 VERSION=3.1 CAPMOD=2 MOBMOD=1
+TOX=310E-10 XJ=9.0E-7 NCH=8.2E16 VTH0=1.026
+K1=1.724 K2=-0.1212 K3=0 K3B=0 WO=2.5E-6 NLX=4.80E-9
+DVT0W=0 DVT1W=0 DVT2W=-0.032 DVT0=0.1466 DVT1=0.038 DVT2=0.1394
+U0=687.22 UA=2.34E-9 UB=-1.85E-18 UC=-1.29E-11 VSAT=1.64E5
+A0=0.4453 AGS=0 B0=0 B1=0 KETA=-0.0569 A1=0 A2=1
+RDSW=376.9 PRWG=0 PRWB=0 WR=1 WINT=2.58E-8 LINT=1.86E-8
+XL=0 XW=0 DWG=0 DWB=0 VOFF=-0.1056 NFACTOR=0.8025
+CIT=0 CDSC=-2.59E-5 CDSCD=0 CDSCB=0 ETA0=0 ETAB=0
+DSUB=0.0117 PCLM=0.6184 PDIBLC1=0.0251 PDIBLC2=0.00202 PDIBLCB=0
+DROUT=0.0772 PSCBE1=2.77E9 PSCBE2=3.11E-8 PVAG=0 DELTA=0.01 PRT=0
+UTE=-1.5 KT1=0 KT1L=0 KT2=0 UA1=4.3E-9 UB1=-7.6E-18
+UC1=-5.6E-11 AT=3.3E4 WL=0 WLN=1 WW=0 WWN=1
+WWL=0 LL=0 LLN=1 LW=0 LWN=1 LWL=0
+XPART=0 +CGD0=1.99E-10 CGS0=1.99E-10 CGB0=5.75E-10 CJ=4.23E-4
+PB=0.99 MJ=0.4496 CJSW=3.83 PBSW=0.1083 MJSW=0.1084
+PVTH0=0.02128 PRDSW=-16.155 PK2=0.0253 WKETA=0.01886 LKETA=0.0205)
```


UTMOST GENERATED SPICE DECK FROM MEASURED SMFL CMOS PROCESS DEVICE CHARACTERISTICS

*1-15-2007 FROM ROB SAXER UTMOST EXTRACTIONS

```
.MODEL RITSMFLP49 PMOS (LEVEL=49 VERSION=3.1 CAPMOD=2 MOBMOD=1
+TOX=310E-10 XJ=8.8E-7 NCH=3.1E16 VTH0=-1.166
+K1=0.3029 K2=0.1055 K3=0 K3B=0 WO=2.5E-6 NLX=2.01E-8
+DVT0W=0 DVT1W=0 DVT2W=-0.032 DVT0=2 DVT1=0.5049 DVT2=-0.0193
+U0=232.53 UA=4E-9 UB=-2.26E-18 UC=-6.80E-11 VSAT=4.40E4
+A0=0.6045 AGS=0 B0=0 B1=0 KETA=-0.0385 A1=0 A2=1
+RDSW=1230 PRWG=0 PRWB=0 WR=1 WINT=1.67E-8 LINT=6.50E-8
+XL=0 XW=0 DWG=0 DWB=0 VOFF=-0.0619 NFACTOR=1.454
+CIT=0 CDSC=-4.30E-4 CDSCD=0 CDSCB=0 ETA0=0 ETAB=0
+DSUB=0.2522 PCLM=5.046 PDIBLC1=0 PDIBLC2=1E-5 PDIBLCB=0
+DROUT=0.2522 PSCBE1=2.8E9 PSCBE2=2.98E-8 PVAG=0 DELTA=0.01 PRT=0
+UTE=-1.5 KT1=0 KT1L=0 KT2=0 UA1=4.3E-9 UB1=-7.6E-18
+UC1=-5.6E-11 AT=3.3E4 WL=0 WLN=1 WW=0 WWN=1
+WWL=0 LL=0 LLN=1 LW=0 LWN=1 LWL=0
+XPART=0 +CGD0=2.4E-10 CGS0=2.4E-10 CGB0=5.75E-10 CJ=7.27E-4
+PB=0.97 MJ=0.496 CJSW=3.115 PBSW=0.99 MJSW=0.2654
+PVTH0=0.00942 PRDSW=-231.3 PK2=1.397 WKETA=1.863 LKETA=5.729)
```

SMFL CMOS PROCESS "HOT & COLD" SPICE MODELS

All parameters the same except those listed are changed to give more transistor current for the hot models:

```
.model hot nmos ( LEVEL = 11    VERSION = 3.1
TOX = 2.70E-8    VTH0= 0.926    U0 = 750    RDSW = 330)
.model hot pmos ( LEVEL = 11    VERSION = 3.1
TOX = 2.70E-8    VTH0= -1.066    U0 = 250    RDSW = 1.00E3)

.model cold nmos ( LEVEL = 11    VERSION = 3.1
TOX = 3.50E-8    VTH0= 1.126    U0 = 620    RDSW = 410)
.model cold pmos ( LEVEL = 11    VERSION = 3.1
TOX = 3.50E-8    VTH0= -1.266    U0 = 200    RDSW = 1.45E3)
```

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HOMWORK – SPICE MOSFET MODELS

1. Write an abstract that summarizes the main ideas presented in this document.

